

INDUCTANCE

30.1. IDENTIFY and SET UP: Apply Eq.(30.4).

EXECUTE: (a) $|\mathcal{E}_2| = M \left| \frac{di_1}{dt} \right| = (3.25 \times 10^{-4} \text{ H})(830 \text{ A/s}) = 0.270 \text{ V}$; yes, it is constant.

(b) $|\mathcal{E}_1| = M \left| \frac{di_2}{dt} \right|$; M is a property of the pair of coils so is the same as in part (a). Thus $|\mathcal{E}_1| = 0.270 \text{ V}$.

EVALUATE: The induced emf is the same in either case. A constant di/dt produces a constant emf.

30.2. IDENTIFY: $\mathcal{E}_1 = M \left| \frac{\Delta i_2}{\Delta t} \right|$ and $\mathcal{E}_2 = M \left| \frac{\Delta i_1}{\Delta t} \right|$. $M = \left| \frac{N_2 \Phi_{B2}}{i_1} \right|$, where Φ_{B2} is the flux through one turn of the second coil.

SET UP: M is the same whether we consider an emf induced in coil 1 or in coil 2.

EXECUTE: (a) $M = \frac{\mathcal{E}_2}{|\Delta i_1 / \Delta t|} = \frac{1.65 \times 10^{-3} \text{ V}}{0.242 \text{ A/s}} = 6.82 \times 10^{-3} \text{ H} = 6.82 \text{ mH}$

(b) $\Phi_{B2} = \frac{M i_1}{N_2} = \frac{(6.82 \times 10^{-3} \text{ H})(1.20 \text{ A})}{25} = 3.27 \times 10^{-4} \text{ Wb}$

(c) $\mathcal{E}_1 = M \left| \frac{\Delta i_2}{\Delta t} \right| = (6.82 \times 10^{-3} \text{ H})(0.360 \text{ A/s}) = 2.46 \times 10^{-3} \text{ V} = 2.46 \text{ mV}$

EVALUATE: We can express M either in terms of the total flux through one coil produced by a current in the other coil, or in terms of the emf induced in one coil by a changing current in the other coil.

30.3. IDENTIFY: Replace units of Wb, A and Ω by their equivalents.

SET UP: $1 \text{ Wb} = 1 \text{ T} \cdot \text{m}^2$. $1 \text{ T} = 1 \text{ N}/(\text{A} \cdot \text{m})$. $1 \text{ N} \cdot \text{m} = 1 \text{ J}$. $1 \text{ A} = 1 \text{ C/s}$. $1 \text{ V} = 1 \text{ J/C}$. $1 \text{ V/A} = 1 \Omega$.

EXECUTE: $1 \text{ H} = 1 \text{ Wb/A} = 1 \text{ T} \cdot \text{m}^2 / \text{A} = 1 \text{ N} \cdot \text{m} / \text{A}^2 = 1 \text{ J/A}^2 = 1 (\text{J}/[\text{A} \cdot \text{C}])\text{s} = 1 (\text{V/A})\text{s} = 1 \Omega \cdot \text{s}$.

EVALUATE: We may use whichever equivalent unit is the most convenient in a particular problem.

30.4. IDENTIFY: Changing flux from one object induces an emf in another object.

(a) **SET UP:** The magnetic field due to a solenoid is $B = \mu_0 n I$.

EXECUTE: The above formula gives

$$B_1 = \frac{(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})(300)(0.120 \text{ A})}{0.250 \text{ m}} = 1.81 \times 10^{-4} \text{ T}$$

The average flux through each turn of the inner solenoid is therefore

$$\Phi_B = B_1 A = (1.81 \times 10^{-4} \text{ T})\pi(0.0100 \text{ m})^2 = 5.68 \times 10^{-8} \text{ Wb}$$

(b) **SET UP:** The flux is the same through each turn of both solenoids due to the geometry, so

$$M = \frac{N_2 \Phi_{B,2}}{i_1} = \frac{N_2 \Phi_{B,1}}{i_1}$$

EXECUTE: $M = \frac{(25)(5.68 \times 10^{-8} \text{ Wb})}{0.120 \text{ A}} = 1.18 \times 10^{-5} \text{ H}$

(c) **SET UP:** The induced emf is $\mathcal{E}_2 = -M \frac{di_1}{dt}$.

EXECUTE: $\mathcal{E}_2 = -(1.18 \times 10^{-5} \text{ H})(1750 \text{ A/s}) = -0.0207 \text{ V}$

EVALUATE: A mutual inductance around 10^{-5} H is not unreasonable.

30.5. IDENTIFY and SET UP: Apply Eq.(30.5).

EXECUTE: (a) $M = \frac{N_2 \Phi_{B2}}{i_1} = \frac{400(0.0320 \text{ Wb})}{6.52 \text{ A}} = 1.96 \text{ H}$

(b) $M = \frac{N_1 \Phi_{B1}}{i_2}$ so $\Phi_{B1} = \frac{Mi_2}{N_1} = \frac{(1.96 \text{ H})(2.54 \text{ A})}{700} = 7.11 \times 10^{-3} \text{ Wb}$

EVALUATE: M relates the current in one coil to the flux through the other coil. Eq.(30.5) shows that M is the same for a pair of coils, no matter which one has the current and which one has the flux.

30.6. IDENTIFY: A changing current in an inductor induces an emf in it.

(a) **SET UP:** The self-inductance of a toroidal solenoid is $L = \frac{\mu_0 N^2 A}{2\pi r}$.

EXECUTE: $L = \frac{(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})(500)^2 (6.25 \times 10^{-4} \text{ m}^2)}{2\pi(0.0400 \text{ m})} = 7.81 \times 10^{-4} \text{ H}$

(b) **SET UP:** The magnitude of the induced emf is $\mathcal{E} = L \frac{di}{dt}$.

EXECUTE: $\mathcal{E} = (7.81 \times 10^{-4} \text{ H}) \left(\frac{5.00 \text{ A} - 2.00 \text{ A}}{3.00 \times 10^{-3} \text{ s}} \right) = 0.781 \text{ V}$

(c) The current is decreasing, so the induced emf will be in the same direction as the current, which is from a to b , making b at a higher potential than a .

EVALUATE: This is a reasonable value for self-inductance, in the range of a mH.

30.7. IDENTIFY: $\mathcal{E} = L \left| \frac{\Delta i}{\Delta t} \right|$ and $L = \frac{N\Phi_B}{i}$.

SET UP: $\frac{\Delta i}{\Delta t} = 0.0640 \text{ A/s}$

EXECUTE: (a) $L = \frac{\mathcal{E}}{|\Delta i / \Delta t|} = \frac{0.0160 \text{ V}}{0.0640 \text{ A/s}} = 0.250 \text{ H}$

(b) The average flux through each turn is $\Phi_B = \frac{Li}{N} = \frac{(0.250 \text{ H})(0.720 \text{ A})}{400} = 4.50 \times 10^{-4} \text{ Wb}$.

EVALUATE: The self-induced emf depends on the rate of change of flux and therefore on the rate of change of the current, not on the value of the current.

30.8. IDENTIFY: Combine the two expressions for L : $L = N\Phi_B/i$ and $L = \mathcal{E}/(di/dt)$.

SET UP: Φ_B is the average flux through one turn of the solenoid.

EXECUTE: Solving for N we have $N = \mathcal{E}i/\Phi_B(di/dt) = \frac{(12.6 \times 10^{-3} \text{ V})(1.40 \text{ A})}{(0.00285 \text{ Wb})(0.0260 \text{ A/s})} = 238 \text{ turns}$.

EVALUATE: The induced emf depends on the time rate of change of the total flux through the solenoid.

30.9. IDENTIFY and SET UP: Apply $|\mathcal{E}| = L|di/dt|$. Apply Lenz's law to determine the direction of the induced emf in the coil.

EXECUTE: (a) $|\mathcal{E}| = L(di/dt) = (0.260 \text{ H})(0.0180 \text{ A/s}) = 4.68 \times 10^{-3} \text{ V}$

(b) Terminal a is at a higher potential since the coil pushes current through from b to a and if replaced by a battery it would have the + terminal at a .

EVALUATE: The induced emf is directed so as to oppose the decrease in the current.

30.10. IDENTIFY: Apply $\mathcal{E} = -L \frac{di}{dt}$.

SET UP: The induced emf points from low potential to high potential across the inductor.

EXECUTE: (a) The induced emf points from b to a , in the direction of the current. Therefore, the current is decreasing and the induced emf is directed to oppose this decrease.

(b) $|\mathcal{E}| = L|\Delta i / \Delta t|$, so $|\Delta i / \Delta t| = V_{ab}/L = (1.04 \text{ V})/(0.260 \text{ H}) = 4.00 \text{ A/s}$. In 2.00 s the decrease in i is 8.00 A and the current at 2.00 s is 12.0 A - 8.0 A = 4.0 A.

EVALUATE: When the current is decreasing the end of the inductor where the current enters is at the lower potential. This agrees with our result and with Figure 30.6d in the textbook.

30.11. IDENTIFY and SET UP: Use Eq.(30.6) to relate L to the flux through each turn of the solenoid. Use Eq.(28.23) for the magnetic field through the solenoid.

EXECUTE: $L = \frac{N\Phi_B}{i}$. If the magnetic field is uniform inside the solenoid $\Phi_B = BA$. From Eq.(28.23),

$$B = \mu_0 ni = \mu_0 \left(\frac{N}{l} \right) i \text{ so } \Phi_B = \frac{\mu_0 NiA}{l}. \text{ Then } L = \frac{N \left(\frac{\mu_0 NiA}{l} \right)}{i} = \frac{\mu_0 N^2 A}{l}.$$

EVALUATE: Our result is the same as L for a toroidal solenoid calculated in Example 30.3, except that the average circumference $2\pi r$ of the toroid is replaced by the length l of the straight solenoid.

30.12. IDENTIFY and SET UP: The stored energy is $U = \frac{1}{2}LI^2$. The rate at which thermal energy is developed is $P = I^2R$.

EXECUTE: (a) $U = \frac{1}{2}LI^2 = \frac{1}{2}(12.0 \text{ H})(0.300 \text{ A})^2 = 0.540 \text{ J}$

(b) $P = I^2R = (0.300 \text{ A})^2(180 \Omega) = 16.2 \text{ W} = 16.2 \text{ J/s}$

EVALUATE: (c) No. If I is constant then the stored energy U is constant. The energy being consumed by the resistance of the inductor comes from the emf source that maintains the current; it does not come from the energy stored in the inductor.

30.13. IDENTIFY and SET UP: Use Eq.(30.9) to relate the energy stored to the inductance. Example 30.3 gives the inductance of a toroidal solenoid to be $L = \frac{\mu_0 N^2 A}{2\pi r}$, so once we know L we can solve for N .

EXECUTE: $U = \frac{1}{2}LI^2$ so $L = \frac{2U}{I^2} = \frac{2(0.390 \text{ J})}{(12.0 \text{ A})^2} = 5.417 \times 10^{-3} \text{ H}$

$$N = \sqrt{\frac{2\pi rL}{\mu_0 A}} = \sqrt{\frac{2\pi(0.150 \text{ m})(5.417 \times 10^{-3} \text{ H})}{(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})(5.00 \times 10^{-4} \text{ m}^2)}} = 2850.$$

EVALUATE: L and hence U increase according to the square of N .

30.14. IDENTIFY: A current-carrying inductor has a magnetic field inside of itself and hence stores magnetic energy.

(a) **SET UP:** The magnetic field inside a toroidal solenoid is $B = \frac{\mu_0 NI}{2\pi r}$.

EXECUTE: $B = \frac{\mu_0(300)(5.00 \text{ A})}{2\pi(0.120 \text{ m})} = 2.50 \times 10^{-3} \text{ T} = 2.50 \text{ mT}$

(b) **SET UP:** The self-inductance of a toroidal solenoid is $L = \frac{\mu_0 N^2 A}{2\pi r}$.

EXECUTE: $L = \frac{(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})(300)^2(4.00 \times 10^{-4} \text{ m}^2)}{2\pi(0.0120 \text{ m})} = 6.00 \times 10^{-5} \text{ H}$

(c) **SET UP:** The energy stored in an inductor is $U_L = \frac{1}{2}LI^2$.

EXECUTE: $U_L = \frac{1}{2}(6.00 \times 10^{-5} \text{ H})(5.00 \text{ A})^2 = 7.50 \times 10^{-4} \text{ J}$

(d) **SET UP:** The energy density in a magnetic field is $u = \frac{B^2}{2\mu_0}$.

EXECUTE: $u = \frac{(2.50 \times 10^{-3} \text{ T})^2}{2(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})} = 2.49 \text{ J/m}^3$

(e) $u = \frac{\text{energy}}{\text{volume}} = \frac{\text{energy}}{2\pi rA} = \frac{7.50 \times 10^{-4} \text{ J}}{2\pi(0.120 \text{ m})(4.00 \times 10^{-4} \text{ m}^2)} = 2.49 \text{ J/m}^3$

EVALUATE: An inductor stores its energy in the magnetic field inside of it.

30.15. IDENTIFY: A current-carrying inductor has a magnetic field inside of itself and hence stores magnetic energy.

(a) **SET UP:** The magnetic field inside a solenoid is $B = \mu_0 nI$.

EXECUTE: $B = \frac{(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})(400)(80.0 \text{ A})}{0.250 \text{ m}} = 0.161 \text{ T}$

(b) **SET UP:** The energy density in a magnetic field is $u = \frac{B^2}{2\mu_0}$.

EXECUTE: $u = \frac{(0.161 \text{ T})^2}{2(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})} = 1.03 \times 10^4 \text{ J/m}^3$

(c) **SET UP:** The total stored energy is $U = uV$.

EXECUTE: $U = uV = u(lA) = (1.03 \times 10^4 \text{ J/m}^3)(0.250 \text{ m})(0.500 \times 10^{-4} \text{ m}^2) = 0.129 \text{ J}$

(d) **SET UP:** The energy stored in an inductor is $U = \frac{1}{2}LI^2$.

EXECUTE: Solving for L and putting in the numbers gives

$$L = \frac{2U}{I^2} = \frac{2(0.129 \text{ J})}{(80.0 \text{ A})^2} = 4.02 \times 10^{-5} \text{ H}$$

EVALUATE: An inductor stores its energy in the magnetic field inside of it.

30.16. IDENTIFY: Energy = Pt . $U = \frac{1}{2}LI^2$.

SET UP: $P = 200 \text{ W} = 200 \text{ J/s}$

EXECUTE: (a) Energy = $(200 \text{ W})(24 \text{ h})(3600 \text{ s/h}) = 1.73 \times 10^7 \text{ J}$

$$(b) L = \frac{2U}{I^2} = \frac{2(1.73 \times 10^7 \text{ J})}{(80.0 \text{ A})^2} = 5.41 \times 10^3 \text{ H}$$

EVALUATE: A large value of L and a large current would be required, just for one light bulb. Also, the resistance of the inductor would have to be very small, to avoid a large $P = I^2R$ rate of electrical energy loss.

30.17. IDENTIFY and SET UP: Starting with Eq. (30.9), follow exactly the same steps as in the text except that the magnetic permeability μ is used in place of μ_0 .

EXECUTE: Using $L = \frac{\mu N^2 A}{2\pi r}$ and $B = \frac{\mu NI}{2\pi r}$ gives $u = \frac{B^2}{2\mu}$.

EVALUATE: For a given value of B , the energy density is less when μ is larger than μ_0 .

30.18. IDENTIFY and SET UP: The energy density (energy per unit volume) in a magnetic field (in vacuum) is given by

$$u = \frac{U}{V} = \frac{B^2}{2\mu_0} \text{ (Eq. 30.10).}$$

EXECUTE: (a) $V = \frac{2\mu_0 U}{B^2} = \frac{2(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})(3.60 \times 10^6 \text{ J})}{(0.600 \text{ T})^2} = 25.1 \text{ m}^3$.

$$(b) u = \frac{U}{V} = \frac{B^2}{2\mu_0}$$

$$B = \sqrt{\frac{2\mu_0 U}{V}} = \sqrt{\frac{2(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})(3.60 \times 10^6 \text{ J})}{(0.400 \text{ m})^3}} = 11.9 \text{ T}$$

EVALUATE: Large-scale energy storage in a magnetic field is not practical. The volume in part (a) is quite large and the field in part (b) would be very difficult to achieve.

30.19. IDENTIFY: Apply Kirchoff's loop rule to the circuit. $i(t)$ is given by Eq.(30.14).

SET UP: The circuit is sketched in Figure 30.19.

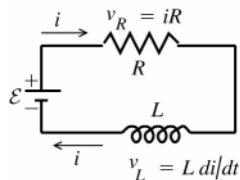


Figure 30.19

$\frac{di}{dt}$ is positive as the current increases from its initial value of zero.

EXECUTE: $\mathcal{E} - v_R - v_L = 0$

$$\mathcal{E} - iR - L \frac{di}{dt} = 0 \text{ so } i = \frac{\mathcal{E}}{R} (1 - e^{-(R/L)t})$$

(a) Initially ($t = 0$), $i = 0$ so $\mathcal{E} - L \frac{di}{dt} = 0$

$$\frac{di}{dt} = \frac{\mathcal{E}}{L} = \frac{6.00 \text{ V}}{2.50 \text{ H}} = 2.40 \text{ A/s}$$

(b) $\mathcal{E} - iR - L \frac{di}{dt} = 0$ (Use this equation rather than Eq.(30.15) since i rather than t is given.)

$$\text{Thus } \frac{di}{dt} = \frac{\mathcal{E} - iR}{L} = \frac{6.00 \text{ V} - (0.500 \text{ A})(8.00 \Omega)}{2.50 \text{ H}} = 0.800 \text{ A/s}$$

$$(c) i = \frac{\mathcal{E}}{R} (1 - e^{-(R/L)t}) = \left(\frac{6.00 \text{ V}}{8.00 \Omega} \right) (1 - e^{-(8.00 \Omega / 2.50 \text{ H})(0.250 \text{ s})}) = 0.750 \text{ A} (1 - e^{-0.800}) = 0.413 \text{ A}$$

(d) Final steady state means $t \rightarrow \infty$ and $\frac{di}{dt} \rightarrow 0$, so $\mathcal{E} - iR = 0$.

$$i = \frac{\mathcal{E}}{R} = \frac{6.00 \text{ V}}{8.00 \Omega} = 0.750 \text{ A}$$

EVALUATE: Our results agree with Fig.30.12 in the textbook. The current is initially zero and increases to its final value of \mathcal{E}/R . The slope of the current in the figure, which is di/dt , decreases with t .

30.20. IDENTIFY: The current decays exponentially.

SET UP: After opening the switch, the current is $i = I_0 e^{-t/RL}$, and the time constant is $\tau = L/R$.

EXECUTE: (a) The initial current is $I_0 = (6.30 \text{ V})/(15.0 \Omega) = 0.420 \text{ A}$. Now solve for L and put in the numbers.

$$L = \frac{-tR}{\ln(i/I_0)} = \frac{-(2.00 \text{ ms})(15.0 \Omega)}{\ln\left(\frac{0.210 \text{ A}}{0.420 \text{ A}}\right)} = 43.3 \text{ mH}$$

(b) $\tau = L/R = (43.3 \text{ mH})/(15.0 \Omega) = 2.89 \text{ ms}$

(c) Solve $i = I_0 e^{-t/\tau}$ for t , giving $t = -\tau \ln(i/I_0) = -(2.89 \text{ ms}) \ln(0.0100) = 13.3 \text{ ms}$.

EVALUATE: In less than 5 time constants, the current is only 1% of its initial value.

30.21. IDENTIFY: $i = \mathcal{E}/R(1 - e^{-t/\tau})$, with $\tau = L/R$. The energy stored in the inductor is $U = \frac{1}{2} Li^2$.

SET UP: The maximum current occurs after a long time and is equal to \mathcal{E}/R .

EXECUTE: (a) $i_{\max} = \mathcal{E}/R$ so $i = i_{\max}/2$ when $(1 - e^{-t/\tau}) = \frac{1}{2}$ and $e^{-t/\tau} = \frac{1}{2}$. $-t/\tau = \ln(\frac{1}{2})$.

$$t = \frac{L \ln 2}{R} = \frac{(\ln 2)(1.25 \times 10^{-3} \text{ H})}{50.0 \Omega} = 17.3 \mu\text{s}$$

(b) $U = \frac{1}{2} U_{\max}$ when $i = i_{\max}/\sqrt{2}$. $1 - e^{-t/\tau} = 1/\sqrt{2}$, so $e^{-t/\tau} = 1 - 1/\sqrt{2} = 0.2929$. $t = -L \ln(0.2929)/R = 30.7 \mu\text{s}$.

EVALUATE: $\tau = L/R = 2.50 \times 10^{-5} \text{ s} = 25.0 \mu\text{s}$. The time in part (a) is 0.692τ and the time in part (b) is 1.23τ .

30.22. IDENTIFY: With S_1 closed and S_2 open, $i(t)$ is given by Eq.(30.14). With S_1 open and S_2 closed, $i(t)$ is given by Eq.(30.18).

SET UP: $U = \frac{1}{2} Li^2$. After S_1 has been closed a long time, i has reached its final value of $I = \mathcal{E}/R$.

EXECUTE: (a) $U = \frac{1}{2} LI^2$ and $I = \sqrt{\frac{2U}{L}} = \sqrt{\frac{2(0.260 \text{ J})}{0.115 \text{ H}}} = 2.13 \text{ A}$. $\mathcal{E} = IR = (2.13 \text{ A})(120 \Omega) = 256 \text{ V}$.

(b) $i = I e^{-(R/L)t}$ and $U = \frac{1}{2} Li^2 = \frac{1}{2} LI^2 e^{-2(R/L)t} = \frac{1}{2} U_0 = \frac{1}{2} (\frac{1}{2} LI^2)$. $e^{-2(R/L)t} = \frac{1}{2}$, so

$$t = -\frac{L}{2R} \ln\left(\frac{1}{2}\right) = -\frac{0.115 \text{ H}}{2(120 \Omega)} \ln\left(\frac{1}{2}\right) = 3.32 \times 10^{-4} \text{ s}$$

EVALUATE: $\tau = L/R = 9.58 \times 10^{-4} \text{ s}$. The time in part (b) is $\tau \ln(2)/2 = 0.347\tau$.

30.23. IDENTIFY: L has units of H and R has units of Ω .

SET UP: $1 \text{ H} = 1 \Omega \cdot \text{s}$

EXECUTE: Units of $L/R = \text{H}/\Omega = (\Omega \cdot \text{s})/\Omega = \text{s} = \text{units of time}$.

EVALUATE: $Rt/L = t/\tau$ is dimensionless.

30.24. IDENTIFY: Apply the loop rule.

SET UP: In applying the loop rule, go around the circuit in the direction of the current. The voltage across the inductor is $-L di/dt$.

EXECUTE: $-L di/dt - iR = 0$. $\frac{di}{dt} = -i \frac{R}{L}$ gives $\int_{I_0}^i \frac{di'}{i'} = -\frac{R}{L} \int_0^t dt'$ and $\ln(i/I_0) = -\frac{R}{L} t$. $i = I_0 e^{-t(R/L)}$.

EVALUATE: di/dt is negative, so there is a potential rise across the inductor; point c is at higher potential than point b . There is a potential drop across the resistor.

30.25. IDENTIFY: Apply the concepts of current decay in an R - L circuit. Apply the loop rule to the circuit. $i(t)$ is given by Eq.(30.18). The voltage across the resistor depends on i and the voltage across the inductor depends on di/dt .

SET UP: The circuit with S_1 closed and S_2 open is sketched in Figure 30.25a.

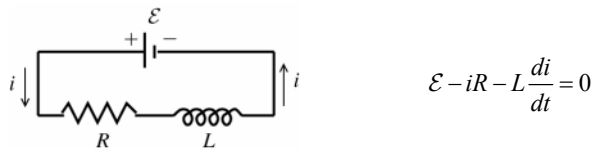
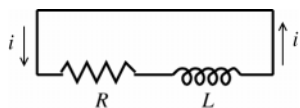


Figure 30.25a

Constant current established means $\frac{di}{dt} = 0$.

EXECUTE: $i = \frac{\mathcal{E}}{R} = \frac{60.0 \text{ V}}{240 \Omega} = 0.250 \text{ A}$

(a) **SET UP:** The circuit with S_2 closed and S_1 open is shown in Figure 30.25b.



$$i = I_0 e^{-(R/L)t}$$

$$\text{At } t = 0, i = I_0 = 0.250 \text{ A}$$

Figure 30.25b

The inductor prevents an instantaneous change in the current; the current in the inductor just after S_2 is closed and S_1 is opened equals the current in the inductor just before this is done.

(b) **EXECUTE:** $i = I_0 e^{-(R/L)t} = (0.250 \text{ A})e^{-(240 \Omega / 0.160 \text{ H})(4.00 \times 10^{-4} \text{ s})} = (0.250 \text{ A})e^{-0.600} = 0.137 \text{ A}$

(c) **SET UP:** See Figure 30.25c.

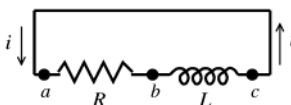


Figure 30.25c

EXECUTE: If we trace around the loop in the direction of the current the potential falls as we travel through the resistor so it must rise as we pass through the inductor: $v_{ab} > 0$ and $v_{bc} < 0$. So point c is at higher potential than point b .

$$v_{ab} + v_{bc} = 0 \text{ and } v_{bc} = -v_{ab}$$

$$\text{Or, } v_{cb} = v_{ab} = iR = (0.137 \text{ A})(240 \Omega) = 32.9 \text{ V}$$

(d) $i = I_0 e^{-(R/L)t}$

$$i = \frac{1}{2} I_0 \text{ says } \frac{1}{2} I_0 = I_0 e^{-(R/L)t} \text{ and } \frac{1}{2} = e^{-(R/L)t}$$

Taking natural logs of both sides of this equation gives $\ln(\frac{1}{2}) = -Rt/L$

$$t = \left(\frac{0.160 \text{ H}}{240 \Omega} \right) \ln 2 = 4.62 \times 10^{-4} \text{ s}$$

EVALUATE: The current decays, as shown in Fig. 30.13 in the textbook. The time constant is $\tau = L/R = 6.67 \times 10^{-4} \text{ s}$. The values of t in the problem are less than one time constant. At any instant the potential drop across the resistor (in the direction of the current) equals the potential rise across the inductor.

30.26. **IDENTIFY:** Apply Eq.(30.14).

SET UP: $v_{ab} = iR$. $v_{bc} = L \frac{di}{dt}$. The current is increasing, so di/dt is positive.

EXECUTE: (a) At $t = 0$, $i = 0$. $v_{ab} = 0$ and $v_{bc} = 60 \text{ V}$.

(b) As $t \rightarrow \infty$, $i \rightarrow \mathcal{E}/R$ and $di/dt \rightarrow 0$. $v_{ab} \rightarrow 60 \text{ V}$ and $v_{bc} \rightarrow 0$.

(c) When $i = 0.150 \text{ A}$, $v_{ab} = iR = 36.0 \text{ V}$ and $v_{bc} = 60.0 \text{ V} - 36.0 \text{ V} = 24.0 \text{ V}$.

EVALUATE: At all times, $\mathcal{E} = v_{ab} + v_{bc}$, as required by the loop rule.

30.27. **IDENTIFY:** $i(t)$ is given by Eq.(30.14).

SET UP: The power input from the battery is $\mathcal{E}i$. The rate of dissipation of energy in the resistance is $i^2 R$. The voltage across the inductor has magnitude $L di/dt$, so the rate at which energy is being stored in the inductor is $iL di/dt$.

EXECUTE: (a) $P = \mathcal{E}i = \mathcal{E}I_0(1 - e^{-(R/L)t}) = \frac{\mathcal{E}^2}{R}(1 - e^{-(R/L)t}) = \frac{(6.00 \text{ V})^2}{8.00 \Omega}(1 - e^{-(8.00 \Omega / 2.50 \text{ H})t})$.

$$P = (4.50 \text{ W})(1 - e^{-(3.20 \text{ s}^{-1})t})$$

(b) $P_R = i^2 R = \frac{\mathcal{E}^2}{R}(1 - e^{-(R/L)t})^2 = \frac{(6.00 \text{ V})^2}{8.00 \Omega}(1 - e^{-(8.00 \Omega / 2.50 \text{ H})t})^2 = (4.50 \text{ W})(1 - e^{-(3.20 \text{ s}^{-1})t})^2$

(c) $P_L = iL \frac{di}{dt} = \frac{\mathcal{E}}{R}(1 - e^{-(R/L)t})L \left(\frac{\mathcal{E}}{L} e^{-(R/L)t} \right) = \frac{\mathcal{E}^2}{R}(e^{-(R/L)t} - e^{-2(R/L)t})$

$$P_L = (4.50 \text{ W})(e^{-(3.20 \text{ s}^{-1})t} - e^{-(6.40 \text{ s}^{-1})t})$$

EVALUATE: (d) Note that if we expand the square in part (b), then parts (b) and (c) add to give part (a), and the total power delivered is dissipated in the resistor and inductor. Conservation of energy requires that this be so.

30.28. IDENTIFY: An L - C circuit oscillates, with the energy going back and forth between the inductor and capacitor.

(a) SET UP: The frequency is $f = \frac{\omega}{2\pi}$ and $\omega = \frac{1}{\sqrt{LC}}$, giving $f = \frac{1}{2\pi\sqrt{LC}}$.

EXECUTE: $f = \frac{1}{2\pi\sqrt{(0.280 \times 10^{-3} \text{ H})(20.0 \times 10^{-6} \text{ F})}} = 2.13 \times 10^3 \text{ Hz} = 2.13 \text{ kHz}$

(b) SET UP: The energy stored in a capacitor is $U = \frac{1}{2}CV^2$.

EXECUTE: $U = \frac{1}{2}(20.0 \times 10^{-6} \text{ F})(150.0 \text{ V})^2 = 0.225 \text{ J}$

(c) SET UP: The current in the circuit is $i = -\omega Q \sin \omega t$, and the energy stored in the inductor is $U = \frac{1}{2}Li^2$.

EXECUTE: First find ω and Q . $\omega = 2\pi f = 1.336 \times 10^4 \text{ rad/s}$.

$$Q = CV = (20.0 \times 10^{-6} \text{ F})(150.0 \text{ V}) = 3.00 \times 10^{-3} \text{ C}$$

Now calculate the current:

$$i = -(1.336 \times 10^4 \text{ rad/s})(3.00 \times 10^{-3} \text{ C}) \sin[(1.336 \times 10^4 \text{ rad/s})(1.30 \times 10^{-3} \text{ s})]$$

Notice that the argument of the sine is in *radians*, so convert it to degrees if necessary. The result is $i = -39.92 \text{ A}$

Now find the energy in the inductor: $U = \frac{1}{2}Li^2 = \frac{1}{2}(0.280 \times 10^{-3} \text{ H})(-39.92 \text{ A})^2 = 0.223 \text{ J}$

EVALUATE: At the end of 1.30 ms, nearly all the energy is now in the inductor, leaving very little in the capacitor.

30.29. IDENTIFY: The energy moves back and forth between the inductor and capacitor.

(a) SET UP: The period is $T = \frac{1}{f} = \frac{1}{\omega/2\pi} = \frac{2\pi}{\omega} = 2\pi\sqrt{LC}$.

EXECUTE: Solving for L gives

$$L = \frac{T^2}{4\pi^2 C} = \frac{(8.60 \times 10^{-5} \text{ s})^2}{4\pi^2 (7.50 \times 10^{-9} \text{ F})} = 2.50 \times 10^{-2} \text{ H} = 25.0 \text{ mH}$$

(b) SET UP: The charge on a capacitor is $Q = CV$.

EXECUTE: $Q = CV = (7.50 \times 10^{-9} \text{ F})(12.0 \text{ V}) = 9.00 \times 10^{-8} \text{ C}$

(c) SET UP: The stored energy is $U = Q^2/2C$.

EXECUTE: $U = \frac{(9.00 \times 10^{-8} \text{ C})^2}{2(7.50 \times 10^{-9} \text{ F})} = 5.40 \times 10^{-7} \text{ J}$

(d) SET UP: The maximum current occurs when the capacitor is discharged, so the inductor has all the initial energy. $U_L + U_C = U_{\text{Total}}$. $\frac{1}{2}Li^2 + 0 = U_{\text{Total}}$.

EXECUTE: Solve for the current:

$$I = \sqrt{\frac{2U_{\text{Total}}}{L}} = \sqrt{\frac{2(5.40 \times 10^{-7} \text{ J})}{2.50 \times 10^{-2} \text{ H}}} = 6.58 \times 10^{-3} \text{ A} = 6.58 \text{ mA}$$

EVALUATE: The energy oscillates back and forth forever. However if there is any resistance in the circuit, no matter how small, all this energy will eventually be dissipated as heat in the resistor.

30.30. IDENTIFY: The circuit is described in Figure 30.14 of the textbook.

SET UP: The energy stored in the inductor is $U_L = \frac{1}{2}Li^2$ and the energy stored in the capacitor is $U_C = q^2/2C$. Initially,

$U_C = \frac{1}{2}CV^2$, with $V = 12.0 \text{ V}$. The period of oscillation is $T = 2\pi\sqrt{LC} = 2\pi\sqrt{(12.0 \times 10^{-3} \text{ H})(18.0 \times 10^{-6} \text{ F})} = 2.92 \text{ ms}$.

EXECUTE: **(a)** Energy conservation says $U_L(\text{max}) = U_C(\text{max})$, and $\frac{1}{2}Li_{\text{max}}^2 = \frac{1}{2}CV^2$.

$i_{\text{max}} = V\sqrt{C/L} = (22.5 \text{ V})\sqrt{\frac{18 \times 10^{-6} \text{ F}}{12 \times 10^{-3} \text{ H}}} = 0.871 \text{ A}$. The charge on the capacitor is zero because all the energy is in

the inductor.

(b) From Figure 30.14 in the textbook, $q = 0$ at $t = T/4 = 0.730 \text{ ms}$ and at $t = 3T/4 = 2.19 \text{ ms}$.

(c) $q_0 = CV = (18 \mu\text{F})(22.5 \text{ V}) = 405 \mu\text{C}$ is the maximum charge on the plates. The graphs are sketched in Figure 30.30. q refers to the charge on one plate and the sign of i indicates the direction of the current.

EVALUATE: If the capacitor is fully charged at $t = 0$ it is fully charged again at $t = T/2$, but with the opposite polarity.

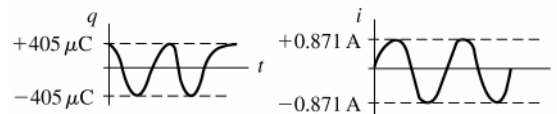


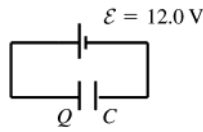
Figure 30.30

30.31. IDENTIFY and SET UP: The angular frequency is given by Eq.(30.22). $q(t)$ and $i(t)$ are given by Eqs.(30.21) and (30.23). The energy stored in the capacitor is $U_C = \frac{1}{2}CV^2 = q^2/2C$. The energy stored in the inductor is $U_L = \frac{1}{2}Li^2$.

EXECUTE: (a) $\omega = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{(1.50 \text{ H})(6.00 \times 10^{-5} \text{ F})}} = 105.4 \text{ rad/s}$, which rounds to 105 rad/s. The period is

given by $T = \frac{2\pi}{\omega} = \frac{2\pi}{105.4 \text{ rad/s}} = 0.0596 \text{ s}$

(b) The circuit containing the battery and capacitor is sketched in Figure 30.31.



$$\mathcal{E} - \frac{Q}{C} = 0$$

$$Q = \mathcal{E}C = (12.0 \text{ V})(6.00 \times 10^{-5} \text{ F}) = 7.20 \times 10^{-4} \text{ C}$$

Figure 30.31

(c) $U = \frac{1}{2}CV^2 = \frac{1}{2}(6.00 \times 10^{-5} \text{ F})(12.0 \text{ V})^2 = 4.32 \times 10^{-3} \text{ J}$

(d) $q = Q \cos(\omega t + \phi)$ (Eq.30.21)

$q = Q$ at $t = 0$ so $\phi = 0$

$q = Q \cos \omega t = (7.20 \times 10^{-4} \text{ C}) \cos([105.4 \text{ rad/s}][0.0230 \text{ s}]) = -5.42 \times 10^{-4} \text{ C}$

The minus sign means that the capacitor has discharged fully and then partially charged again by the current maintained by the inductor; the plate that initially had positive charge now has negative charge and the plate that initially had negative charge now has positive charge.

(e) $i = -\omega Q \sin(\omega t + \phi)$ (Eq.30.23)

$i = -(105 \text{ rad/s})(7.20 \times 10^{-4} \text{ C}) \sin([105.4 \text{ rad/s}][0.0230 \text{ s}]) = -0.050 \text{ A}$

The negative sign means the current is counterclockwise in Figure 30.15 in the textbook.

or

$\frac{1}{2}Li^2 + \frac{q^2}{2C} = \frac{Q^2}{2C}$ gives $i = \pm \sqrt{\frac{1}{LC} \sqrt{Q^2 - q^2}}$ (Eq.30.26)

$i = \pm(105 \text{ rad/s}) \sqrt{(7.20 \times 10^{-4} \text{ C})^2 - (-5.42 \times 10^{-4} \text{ C})^2} = \pm 0.050 \text{ A}$, which checks.

(f) $U_C = \frac{q^2}{2C} = \frac{(-5.42 \times 10^{-4} \text{ C})^2}{2(6.00 \times 10^{-5} \text{ F})} = 2.45 \times 10^{-3} \text{ J}$

$U_L = \frac{1}{2}Li^2 = \frac{1}{2}(1.50 \text{ H})(0.050 \text{ A})^2 = 1.87 \times 10^{-3} \text{ J}$

EVALUATE: Note that $U_C + U_L = 2.45 \times 10^{-3} \text{ J} + 1.87 \times 10^{-3} \text{ J} = 4.32 \times 10^{-3} \text{ J}$.

This agrees with the total energy initially stored in the capacitor, $U = \frac{Q^2}{2C} = \frac{(7.20 \times 10^{-4} \text{ C})^2}{2(6.00 \times 10^{-5} \text{ F})} = 4.32 \times 10^{-3} \text{ J}$.

Energy is conserved. At some times there is energy stored in both the capacitor and the inductor. When $i = 0$ all the energy is stored in the capacitor and when $q = 0$ all the energy is stored in the inductor. But at all times the total energy stored is the same.

30.32. IDENTIFY: $\omega = \frac{1}{\sqrt{LC}} = 2\pi f$

SET UP: ω is the angular frequency in rad/s and f is the corresponding frequency in Hz.

EXECUTE: (a) $L = \frac{1}{4\pi^2 f^2 C} = \frac{1}{4\pi^2 (1.6 \times 10^6 \text{ Hz})^2 (4.18 \times 10^{-12} \text{ F})} = 2.37 \times 10^{-3} \text{ H}$.

(b) The maximum capacitance corresponds to the minimum frequency.

$C_{\text{max}} = \frac{1}{4\pi^2 f_{\text{min}}^2 L} = \frac{1}{4\pi^2 (5.40 \times 10^5 \text{ Hz})^2 (2.37 \times 10^{-3} \text{ H})} = 3.67 \times 10^{-11} \text{ F} = 36.7 \text{ pF}$

EVALUATE: To vary f by a factor of three (approximately the range in this problem), C must be varied by a factor of nine.

30.33. IDENTIFY: Apply energy conservation and Eqs. (30.22) and (30.23).

SET UP: If I is the maximum current, $\frac{1}{2}LI^2 = \frac{Q^2}{2C}$. For the inductor, $U_L = \frac{1}{2}LI^2$.

EXECUTE: (a) $\frac{1}{2}LI^2 = \frac{Q^2}{2C}$ gives $Q = I\sqrt{LC} = (0.750 \text{ A})\sqrt{(0.0800 \text{ H})(1.25 \times 10^{-9} \text{ F})} = 7.50 \times 10^{-6} \text{ C}$.

(b) $\omega = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{(0.0800 \text{ H})(1.25 \times 10^{-9} \text{ F})}} = 1.00 \times 10^5 \text{ rad/s}$. $f = \frac{\omega}{2\pi} = 1.59 \times 10^4 \text{ Hz}$.

(c) $q = Q$ at $t = 0$ means $\phi = 0$. $i = -\omega Q \sin(\omega t)$, so
 $i = -(1.00 \times 10^5 \text{ rad/s})(7.50 \times 10^{-6} \text{ C}) \sin([1.00 \times 10^5 \text{ rad/s}][2.50 \times 10^{-3} \text{ s}]) = -0.7279 \text{ A}$.
 $U_L = \frac{1}{2} Li^2 = \frac{1}{2} (0.0800 \text{ H})(-0.7279 \text{ A})^2 = 0.0212 \text{ J}$.

EVALUATE: The total energy of the system is $\frac{1}{2} LI^2 = 0.0225 \text{ J}$. At $t = 2.50 \text{ ms}$, the current is close to its maximum value and most of the system's energy is stored in the inductor.

30.34. IDENTIFY: Apply Eq.(30.25).

SET UP: $q = Q$ when $i = 0$. $i = i_{\max}$ when $q = 0$. $1/\sqrt{LC} = 1917 \text{ s}^{-1}$.

EXECUTE: (a) $\frac{1}{2} Li_{\max}^2 = \frac{Q^2}{2C}$. $Q = i_{\max} \sqrt{LC} = (0.850 \times 10^{-3} \text{ A}) \sqrt{(0.0850 \text{ H})(3.20 \times 10^{-6} \text{ F})} = 4.43 \times 10^{-7} \text{ C}$

(b) $q = \sqrt{Q^2 - LCi^2} = \sqrt{(4.43 \times 10^{-7} \text{ C})^2 - \left(\frac{5.00 \times 10^{-4} \text{ A}}{1917 \text{ s}^{-1}}\right)^2} = 3.58 \times 10^{-7} \text{ C}$.

EVALUATE: The value of q calculated in part (b) is less than the maximum value Q calculated in part (a).

30.35. IDENTIFY: $q = Q \cos(\omega t + \phi)$ and $i = -\omega Q \sin(\omega t + \phi)$

SET UP: $U_C = \frac{q^2}{2C}$. $U_L = \frac{1}{2} Li^2$.

EXECUTE: (a) $U_C = \frac{1}{2} \frac{q^2}{C} = \frac{1}{2} \frac{Q^2 \cos^2(\omega t + \phi)}{C}$.

$U_L = \frac{1}{2} Li^2 = \frac{1}{2} L \omega^2 Q^2 \sin^2(\omega t + \phi) = \frac{1}{2} \frac{Q^2 \sin^2(\omega t + \phi)}{C}$, since $\omega^2 = \frac{1}{\sqrt{LC}}$.

(b) $U_{\text{Total}} = U_C + U_L = \frac{1}{2} \frac{Q^2}{C} \cos^2(\omega t + \phi) + \frac{1}{2} L \omega^2 Q^2 \sin^2(\omega t + \phi)$

$U_{\text{total}} = \frac{1}{2} \frac{Q^2}{C} \cos^2(\omega t + \phi) + \frac{1}{2} L \left(\frac{1}{LC}\right) Q^2 \sin^2(\omega t + \phi) = \frac{1}{2} \frac{Q^2}{C} (\cos^2(\omega t + \phi) + \sin^2(\omega t + \phi)) = \frac{1}{2} \frac{Q^2}{C}$

U_{Total} is a constant.

EVALUATE: Eqs.(30.21) and (30.23) are consistent with conservation of energy in the L - C circuit.

30.36. IDENTIFY: Evaluate $\frac{d^2q}{dt^2}$ and insert into Eq.(20.20).

SET UP: Equation (30.20) is $\frac{d^2q}{dt^2} + \frac{1}{LC}q = 0$.

EXECUTE: $q = Q \cos(\omega t + \phi) \Rightarrow \frac{dq}{dt} = -\omega Q \sin(\omega t + \phi) \Rightarrow \frac{d^2q}{dt^2} = -\omega^2 Q \cos(\omega t + \phi)$.

$\frac{d^2q}{dt^2} + \frac{1}{LC}q = -\omega^2 Q \cos(\omega t + \phi) + \frac{Q}{LC} \cos(\omega t + \phi) = 0 \Rightarrow \omega^2 = \frac{1}{LC} \Rightarrow \omega = \frac{1}{\sqrt{LC}}$.

EVALUATE: The value of ϕ depends on the initial conditions, the value of q at $t = 0$.

30.37. IDENTIFY: The unit of L is H and the unit of C is F.

SET UP: $C = q/V_{ab}$ says $1 \text{ F} = 1 \text{ C/V}$. $1 \text{ H} = 1 \text{ V} \cdot \text{s/A} = 1 \text{ V} \cdot \text{s}^2/\text{C}$.

EXECUTE: $1 \text{ H} \cdot \text{F} = (1 \text{ V} \cdot \text{s}^2/\text{C})(1 \text{ C/V}) = 1 \text{ s}^2$. Therefore, LC has units of s^2 and \sqrt{LC} has units of s .

EVALUATE: Our result shows that ωt is dimensionless, since $\omega = 1/\sqrt{LC}$.

30.38. IDENTIFY: The presence of resistance in an L - R - C circuit affects the frequency of oscillation and causes the amplitude of the oscillations to decrease over time.

(a) **SET UP:** The frequency of damped oscillations is $\omega' = \sqrt{\frac{1}{LC} - \frac{R^2}{4L^2}}$.

EXECUTE: $\omega' = \sqrt{\frac{1}{(22 \times 10^{-3} \text{ H})(15.0 \times 10^{-9} \text{ F})} - \frac{(75.0 \text{ } \Omega)^2}{4(22 \times 10^{-3} \text{ H})^2}} = 5.5 \times 10^4 \text{ rad/s}$

The frequency f is $f = \frac{\omega}{2\pi} = \frac{5.50 \times 10^4 \text{ rad/s}}{2\pi} = 8.76 \times 10^3 \text{ Hz} = 8.76 \text{ kHz}$.

(b) **SET UP:** The amplitude decreases as $A(t) = A_0 e^{-(R/2L)t}$.

EXECUTE: Solving for t and putting in the numbers gives:

$$t = \frac{-2L \ln(A/A_0)}{R} = \frac{-2(22.0 \times 10^{-3} \text{ H}) \ln(0.100)}{75.0 \text{ } \Omega} = 1.35 \times 10^{-3} \text{ s} = 1.35 \text{ ms}$$

(c) **SET UP:** At critical damping, $R = \sqrt{4L/C}$.

EXECUTE: $R = \sqrt{\frac{4(22.0 \times 10^{-3} \text{ H})}{15.0 \times 10^{-9} \text{ F}}} = 2420 \ \Omega$

EVALUATE: The frequency with damping is almost the same as the resonance frequency of this circuit ($1/\sqrt{LC}$), which is plausible because the $75\text{-}\Omega$ resistance is considerably less than the $2420 \ \Omega$ required for critical damping.

30.39. IDENTIFY: Follow the procedure specified in the problem.

SET UP: Make the substitutions $x \rightarrow q$, $m \rightarrow L$, $b \rightarrow R$, $k \rightarrow \frac{1}{C}$.

EXECUTE: (a) Eq. (13.41): $\frac{d^2x}{dt^2} + \frac{b}{m} \frac{dx}{dt} + \frac{kx}{m} = 0$. This becomes $\frac{d^2q}{dt^2} + \frac{R}{L} \frac{dq}{dt} + \frac{q}{LC} = 0$, which is Eq.(30.27).

(b) Eq. (13.43): $\omega' = \sqrt{\frac{k}{m} - \frac{b^2}{4m^2}}$. This becomes $\omega' = \sqrt{\frac{1}{LC} - \frac{R^2}{4L^2}}$, which is Eq.(30.29).

(c) Eq. (13.42): $x = Ae^{-(b/2m)t} \cos(\omega't + \phi)$. This becomes $q = Ae^{-(R/2L)t} \cos(\omega't + \phi)$, which is Eq.(30.28).

EVALUATE: Equations for the L - R - C circuit and for a damped harmonic oscillator have the same form.

30.40. IDENTIFY: For part (a), evaluate the derivatives as specified in the problem. For part (b) set $q = Q$ in Eq.(30.28) and set $dq/dt = 0$ in the expression for dq/dt .

SET UP: In terms of ω' , Eq.(30.28) is $q(t) = Ae^{-(R/2L)t} \cos(\omega't + \phi)$.

EXECUTE: (a) $q = Ae^{-(R/2L)t} \cos(\omega't + \phi)$. $\frac{dq}{dt} = -A \frac{R}{2L} e^{-(R/2L)t} \cos(\omega't + \phi) - \omega' A e^{-(R/2L)t} \sin(\omega't + \phi)$.

$$\frac{d^2q}{dt^2} = A \left(\frac{R}{2L} \right)^2 e^{-(R/2L)t} \cos(\omega't + \phi) + 2\omega' A \frac{R}{2L} e^{-(R/2L)t} \sin(\omega't + \phi) - \omega'^2 A e^{-(R/2L)t} \cos(\omega't + \phi)$$

$$\frac{d^2q}{dt^2} + \frac{R}{L} \frac{dq}{dt} + \frac{q}{LC} = q \left(\left(\frac{R}{2L} \right)^2 - \omega'^2 - \frac{R^2}{2L^2} + \frac{1}{LC} \right) = 0, \text{ so } \omega'^2 = \frac{1}{LC} - \frac{R^2}{4L^2}.$$

(b) At $t = 0$, $q = Q$, $i = \frac{dq}{dt} = 0$, so $q = A \cos \phi = Q$ and $\frac{dq}{dt} = -\frac{R}{2L} A \cos \phi - \omega' A \sin \phi = 0$. This gives $A = \frac{Q}{\cos \phi}$ and

$$\tan \phi = -\frac{R}{2L\omega'} = -\frac{R}{2L\sqrt{1/LC - R^2/4L^2}}.$$

EVALUATE: If $R = 0$, then $A = Q$ and $\phi = 0$.

30.41. IDENTIFY: Evaluate Eq.(30.29).

SET UP: The angular frequency of the circuit is ω' .

EXECUTE: (a) When $R = 0$, $\omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{(0.450 \text{ H})(2.50 \times 10^{-5} \text{ F})}} = 298 \text{ rad/s}$.

(b) We want $\frac{\omega}{\omega_0} = 0.95$, so $\frac{(1/LC - R^2/4L^2)}{1/LC} = 1 - \frac{R^2C}{4L} = (0.95)^2$. This gives

$$R = \sqrt{\frac{4L}{C}(1 - (0.95)^2)} = \sqrt{\frac{4(0.450 \text{ H})(0.0975)}{(2.50 \times 10^{-5} \text{ F})}} = 83.8 \ \Omega.$$

EVALUATE: When R increases, the angular frequency decreases and approaches zero as $R \rightarrow 2\sqrt{L/C}$.

30.42. IDENTIFY: L has units of H and C has units of F.

SET UP: $1 \text{ H} = 1 \ \Omega \cdot \text{s}$. $C = q/V$ says $1 \text{ F} = 1 \text{ C/V}$. $V = IR$ says $1 \text{ V/A} = 1 \ \Omega$.

EXECUTE: The units of L/C are $\frac{\text{H}}{\text{F}} = \frac{\Omega \cdot \text{s}}{\text{C/V}} = \frac{\Omega \cdot \text{V}}{\text{A}} = \Omega^2$. Therefore, the unit of $\sqrt{L/C}$ is Ω .

EVALUATE: For Eq.(30.28) to be valid, $\frac{1}{LC}$ and $\frac{R^2}{4L^2}$ must have the same units, so R and $\sqrt{L/C}$ must have the same units, and we have shown that this is indeed the case.

30.43. IDENTIFY: The emf \mathcal{E}_2 in solenoid 2 produced by changing current i_1 in solenoid 1 is given by $\mathcal{E}_2 = M \left| \frac{\Delta i_1}{\Delta t} \right|$. The mutual inductance of two solenoids is derived in Example 30.1. For the two solenoids in this problem

$$M = \frac{\mu_0 AN_1 N_2}{l}, \text{ where } A \text{ is the cross-sectional area of the inner solenoid and } l \text{ is the length of the outer solenoid.}$$

SET UP: $\mu_0 = 4\pi \times 10^{-7} \text{ T} \cdot \text{m/A}$. Let the outer solenoid be solenoid 1.

EXECUTE: (a) $M = \frac{(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A})\pi(6.00 \times 10^{-4} \text{ m})^2(6750)(15)}{0.500 \text{ m}} = 2.88 \times 10^{-7} \text{ H} = 0.288 \mu\text{H}$

(b) $\mathcal{E}_2 = \left| \frac{\Delta i_1}{\Delta t} \right| = (2.88 \times 10^{-7} \text{ H})(37.5 \text{ A/s}) = 1.08 \times 10^{-5} \text{ V}$

EVALUATE: If current in the inner solenoid changed at 37.5 A/s, the emf induced in the outer solenoid would be $1.08 \times 10^{-5} \text{ V}$.

30.44. IDENTIFY: Apply $\mathcal{E} = -L \frac{di}{dt}$ and $Li = N\Phi_B$.

SET UP: Φ_B is the flux through one turn.

EXECUTE: (a) $\mathcal{E} = -L \frac{di}{dt} = -(3.50 \times 10^{-3} \text{ H}) \frac{d}{dt} ((0.680 \text{ A}) \cos(\pi t / [0.0250 \text{ s}]))$.

$\mathcal{E} = (3.50 \times 10^{-3} \text{ H})(0.680 \text{ A}) \frac{\pi}{0.0250 \text{ s}} \sin(\pi t / [0.0250 \text{ s}])$. Therefore,

$\mathcal{E}_{\text{max}} = (3.50 \times 10^{-3} \text{ H})(0.680 \text{ A}) \frac{\pi}{0.0250 \text{ s}} = 0.299 \text{ V}$.

(b) $\Phi_{B\text{max}} = \frac{Li_{\text{max}}}{N} = \frac{(3.50 \times 10^{-3} \text{ H})(0.680 \text{ A})}{400} = 5.95 \times 10^{-6} \text{ Wb}$.

(c) $\mathcal{E}(t) = -L \frac{di}{dt} = -(3.50 \times 10^{-3} \text{ H})(0.680 \text{ A})(\pi / 0.0250 \text{ s}) \sin(\pi t / 0.0250 \text{ s})$.

$\mathcal{E}(t) = -(0.299 \text{ V}) \sin((125.6 \text{ s}^{-1})t)$. Therefore, at $t = 0.0180 \text{ s}$,

$\mathcal{E}(0.0180 \text{ s}) = -(0.299 \text{ V}) \sin((125.6 \text{ s}^{-1})(0.0180 \text{ s})) = 0.230 \text{ V}$. The magnitude of the induced emf is 0.230 V.

EVALUATE: The maximum emf is when $i = 0$ and at this instant $\Phi_B = 0$.

30.45. IDENTIFY: $\mathcal{E} = -L \frac{di}{dt}$.

SET UP: During an interval in which the graph of i versus t is a straight line, di/dt is constant and equal to the slope of that line.

EXECUTE: (a) The pattern on the oscilloscope is sketched in Figure 30.45.

EVALUATE: (b) Since the voltage is determined by the derivative of the current, the V versus t graph is indeed proportional to the derivative of the current graph.

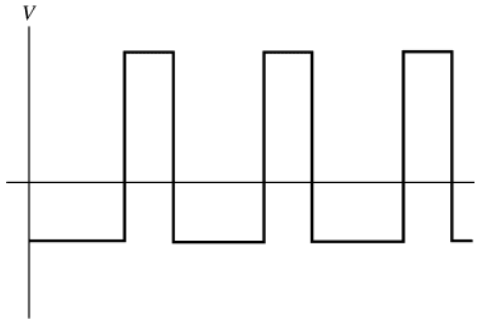


Figure 30.45

30.46. IDENTIFY: Apply $\mathcal{E} = -L \frac{di}{dt}$.

SET UP: $\frac{d}{dt} \cos(\omega t) = -\omega \sin(\omega t)$

EXECUTE: (a) $\mathcal{E} = -L \frac{di}{dt} = -L \frac{d}{dt} ((0.124 \text{ A}) \cos[(240 \pi / \text{s})t])$.

$\mathcal{E} = +(0.250 \text{ H})(0.124 \text{ A})(240 \pi / \text{s}) \sin((240 \pi / \text{s})t) = +(23.4 \text{ V}) \sin((240 \pi / \text{s})t)$.

The graphs are given in Figure 30.46.

(b) $\mathcal{E}_{\text{max}} = 23.4 \text{ V}$; $i = 0$, since the emf and current are 90° out of phase.

(c) $i_{\text{max}} = 0.124 \text{ A}$; $\mathcal{E} = 0$, since the emf and current are 90° out of phase.

EVALUATE: The induced emf depends on the rate at which the current is changing.

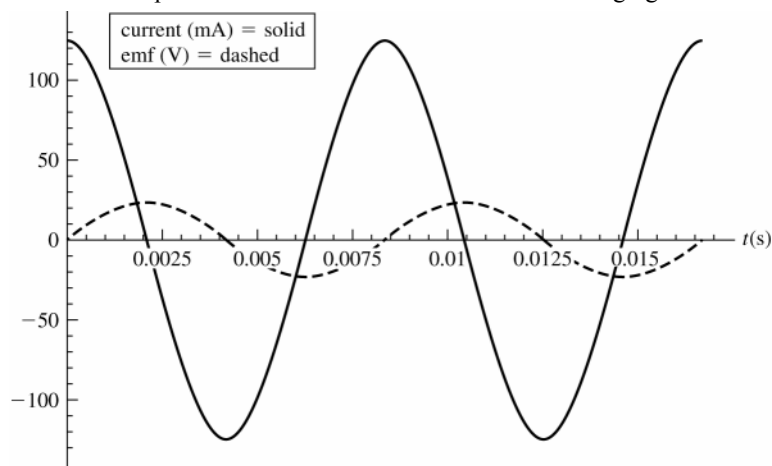


Figure 30.46

30.47. IDENTIFY: Apply $\mathcal{E} = -L \frac{di}{dt}$ to the series and parallel combinations.

SET UP: In series, $i_1 = i_2$ and the voltages add. In parallel the voltages are the same and the currents add.

EXECUTE: (a) Series: $L_1 \frac{di_1}{dt} + L_2 \frac{di_2}{dt} = L_{\text{eq}} \frac{di}{dt}$, but $i_1 = i_2 = i$ for series components so $\frac{di_1}{dt} = \frac{di_2}{dt} = \frac{di}{dt}$ and $L_1 + L_2 = L_{\text{eq}}$.

(b) Parallel: Now $L_1 \frac{di_1}{dt} = L_2 \frac{di_2}{dt} = L_{\text{eq}} \frac{di}{dt}$, where $i = i_1 + i_2$. Therefore, $\frac{di}{dt} = \frac{di_1}{dt} + \frac{di_2}{dt}$. But $\frac{di_1}{dt} = \frac{L_{\text{eq}}}{L_1} \frac{di}{dt}$ and

$$\frac{di_2}{dt} = \frac{L_{\text{eq}}}{L_2} \frac{di}{dt} \cdot \frac{di}{dt} = \frac{L_{\text{eq}}}{L_1} \frac{di}{dt} + \frac{L_{\text{eq}}}{L_2} \frac{di}{dt} \text{ and } L_{\text{eq}} = \left(\frac{1}{L_1} + \frac{1}{L_2} \right)^{-1}.$$

EVALUATE: Inductors in series and parallel combine in the same way as resistors.

30.48. IDENTIFY: Follow the steps outlined in the problem.

SET UP: The energy stored is $U = \frac{1}{2} Li^2$.

EXECUTE: (a) $\oint \vec{B} \cdot d\vec{l} = \mu_0 I_{\text{encl}} \Rightarrow B 2\pi r = \mu_0 i \Rightarrow B = \frac{\mu_0 i}{2\pi r}$.

(b) $d\Phi_B = B dA = \frac{\mu_0 i}{2\pi r} l dr$.

(c) $\Phi_B = \int_a^b d\Phi_B = \frac{\mu_0 i l}{2\pi} \int_a^b \frac{dr}{r} = \frac{\mu_0 i l}{2\pi} \ln(b/a)$.

(d) $L = \frac{N\Phi_B}{i} = l \frac{\mu_0}{2\pi} \ln(b/a)$.

(e) $U = \frac{1}{2} Li^2 = \frac{1}{2} l \frac{\mu_0}{2\pi} \ln(b/a) i^2 = \frac{\mu_0 l i^2}{4\pi} \ln(b/a)$.

EVALUATE: The magnetic field between the conductors is due only to the current in the inner conductor.

30.49. (a) IDENTIFY and SET UP: An end view is shown in Figure 30.49.

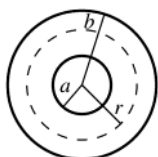


Figure 30.49

Apply Ampere's law to a circular path of radius r .

$$\oint \vec{B} \cdot d\vec{l} = \mu_0 I_{\text{encl}}$$

EXECUTE: $\oint \vec{B} \cdot d\vec{l} = B(2\pi r)$

$I_{\text{encl}} = i$, the current in the inner conductor

Thus $B(2\pi r) = \mu_0 i$ and $B = \frac{\mu_0 i}{2\pi r}$.

(b) IDENTIFY and SET UP: Follow the procedure specified in the problem.

EXECUTE: $u = \frac{B^2}{2\mu_0}$

$dU = u dV$, where $dV = 2\pi r l dr$

$$dU = \frac{1}{2\mu_0} \left(\frac{\mu_0 i}{2\pi r} \right)^2 (2\pi r l) dr = \frac{\mu_0 i^2 l}{4\pi r} dr$$

(c) $U = \int dU = \frac{\mu_0 i^2 l}{4\pi} \int_a^b \frac{dr}{r} = \frac{\mu_0 i^2 l}{4\pi} [\ln r]_a^b$

$$U = \frac{\mu_0 i^2 l}{4\pi} (\ln b - \ln a) = \frac{\mu_0 i^2 l}{4\pi} \ln \left(\frac{b}{a} \right)$$

(d) Eq.(30.9): $U = \frac{1}{2} Li^2$

Part (c): $U = \frac{\mu_0 i^2 l}{4\pi} \ln \left(\frac{b}{a} \right)$

$$\frac{1}{2} Li^2 = \frac{\mu_0 i^2 l}{4\pi} \ln \left(\frac{b}{a} \right)$$

$$L = \frac{\mu_0 l}{2\pi} \ln \left(\frac{b}{a} \right).$$

EVALUATE: The value of L we obtain from these energy considerations agrees with L calculated in part (d) of Problem 30.48 by considering flux and Eq.(30.6)

30.50. IDENTIFY: Apply $L = \frac{N\Phi_B}{i}$ to each solenoid, as in Example 30.3. Use $M = \frac{N_2\Phi_{B2}}{i_1}$ to calculate the mutual

inductance M .

SET UP: The magnetic field produced by solenoid 1 is confined to the space within its windings and is equal to

$$B_1 = \frac{\mu_0 N_1 i_1}{2\pi r}.$$

EXECUTE: **(a)** $L_1 = \frac{N_1\Phi_{B1}}{i_1} = \frac{N_1 A}{i_1} \left(\frac{\mu_0 N_1 i_1}{2\pi r} \right) = \frac{\mu_0 N_1^2 A}{2\pi r}$, $L_2 = \frac{N_2\Phi_{B2}}{i_2} = \frac{N_2 A}{i_2} \left(\frac{\mu_0 N_2 i_2}{2\pi r} \right) = \frac{\mu_0 N_2^2 A}{2\pi r}$.

(b) $M = \frac{N_2 AB_1}{i_1} = \frac{\mu_0 N_1 N_2 A}{2\pi r}$. $M^2 = \left(\frac{\mu_0 N_1 N_2 A}{2\pi r} \right)^2 = \frac{\mu_0 N_1^2 A}{2\pi r} \frac{\mu_0 N_2^2 A}{2\pi r} = L_1 L_2$.

EVALUATE: If the two solenoids are identical, so that $N_1 = N_2$, then $M = L$.

30.51. IDENTIFY: $U = \frac{1}{2} LI^2$. The self-inductance of a solenoid is found in Exercise 30.11 to be $L = \frac{\mu_0 AN^2}{l}$.

SET UP: The length l of the solenoid is the number of turns divided by the turns per unit length.

EXECUTE: **(a)** $L = \frac{2U}{I^2} = \frac{2(10.0 \text{ J})}{(1.50 \text{ A})^2} = 8.89 \text{ H}$

(b) $L = \frac{\mu_0 AN^2}{l}$. If α is the number of turns per unit length, then $N = \alpha l$ and $L = \mu_0 A \alpha^2 l$. For this coil

$$\alpha = 10 \text{ coils/mm} = 10 \times 10^3 \text{ coils/m}. \quad l = \frac{L}{\mu_0 A \alpha^2} = \frac{8.89 \text{ H}}{(4\pi \times 10^{-7} \text{ T} \cdot \text{m/A}) \pi (0.0200 \text{ m})^2 (10 \times 10^3 \text{ coils/m})^2} = 56.3 \text{ m}.$$

This is not a practical length for laboratory use.

EVALUATE: The number of turns is $N = (56.3 \text{ m})(10 \times 10^3 \text{ coils/m}) = 5.63 \times 10^5$ turns. The length of wire in the solenoid is the circumference C of one turn times the number of turns. $C = \pi d = \pi(4.00 \times 10^{-2} \text{ m}) = 0.126 \text{ m}$. The length of wire is $(0.126 \text{ m})(5.63 \times 10^5) = 7.1 \times 10^4 \text{ m} = 71 \text{ km}$. This length of wire will have a large resistance and $I^2 R$ electrical energy losses will be very large.

30.52. IDENTIFY: This is an R - L circuit and $i(t)$ is given by Eq.(30.14).

SET UP: When $t \rightarrow \infty$, $i \rightarrow i_f = V/R$.

EXECUTE: **(a)** $R = \frac{V}{i_f} = \frac{12.0 \text{ V}}{6.45 \times 10^{-3} \text{ A}} = 1860 \Omega$.

(b) $i = i_f(1 - e^{-(R/L)t})$ so $\frac{Rt}{L} = -\ln(1 - i/i_f)$ and $L = \frac{-Rt}{\ln(1 - i/i_f)} = \frac{-(1860 \Omega)(7.25 \times 10^{-4} \text{ s})}{\ln(1 - (4.86/6.45))} = 0.963 \text{ H}$.

EVALUATE: The current after a long time depends only on R and is independent of L . The value of R/L determines how rapidly the final value of i is reached.

30.53. IDENTIFY and SET UP: Follow the procedure specified in the problem. $L = 2.50 \text{ H}$, $R = 8.00 \Omega$,

$$\mathcal{E} = 6.00 \text{ V}. i = (\mathcal{E}/R)(1 - e^{-t/\tau}), \tau = L/R$$

$$\text{EXECUTE: (a) Eq.(30.9): } U_L = \frac{1}{2}Li^2$$

$$t = \tau \text{ so } i = (\mathcal{E}/R)(1 - e^{-1}) = (6.00 \text{ V}/8.00 \Omega)(1 - e^{-1}) = 0.474 \text{ A}$$

$$\text{Then } U_L = \frac{1}{2}Li^2 = \frac{1}{2}(2.50 \text{ H})(0.474 \text{ A})^2 = 0.281 \text{ J}$$

$$\text{Exercise 30.27 (c): } P_L = \frac{dU_L}{dt} = Li \frac{di}{dt}$$

$$i = \left(\frac{\mathcal{E}}{R}\right)(1 - e^{-t/\tau}); \frac{di}{dt} = \left(\frac{\mathcal{E}}{L}\right)e^{-(R/L)t} = \frac{\mathcal{E}}{L}e^{-t/\tau}$$

$$P_L = L \left(\frac{\mathcal{E}}{R}(1 - e^{-t/\tau})\right) \left(\frac{\mathcal{E}}{L}e^{-t/\tau}\right) = \frac{\mathcal{E}^2}{R}(e^{-t/\tau} - e^{-2t/\tau})$$

$$U_L = \int_0^\tau P_L dt = \frac{\mathcal{E}^2}{R} \int_0^\tau (e^{-t/\tau} - e^{-2t/\tau}) dt = \frac{\mathcal{E}^2}{R} \left[-\tau e^{-t/\tau} + \frac{\tau}{2} e^{-2t/\tau} \right]_0^\tau$$

$$U_L = -\frac{\mathcal{E}^2}{R} \tau \left[e^{-1} - \frac{1}{2} e^{-2} \right]_0^\tau = \frac{\mathcal{E}^2}{R} \tau \left[1 - \frac{1}{2} - e^{-1} + \frac{1}{2} e^{-2} \right]$$

$$U_L = \left(\frac{\mathcal{E}^2}{2R}\right) \left(\frac{L}{R}\right) (1 - 2e^{-1} + e^{-2}) = \frac{1}{2} \left(\frac{\mathcal{E}}{R}\right)^2 L (1 - 2e^{-1} + e^{-2})$$

$$U_L = \frac{1}{2} \left(\frac{6.00 \text{ V}}{8.00 \Omega}\right)^2 (2.50 \text{ H})(0.3996) = 0.281 \text{ J, which checks.}$$

(b) Exercise 30.27(a): The rate at which the battery supplies energy is $P_\mathcal{E} = \mathcal{E}i = \mathcal{E} \left(\frac{\mathcal{E}}{R}(1 - e^{-t/\tau})\right) = \frac{\mathcal{E}^2}{R}(1 - e^{-t/\tau})$

$$U_\mathcal{E} = \int_0^\tau P_\mathcal{E} dt = \frac{\mathcal{E}^2}{R} \int_0^\tau (1 - e^{-t/\tau}) dt = \frac{\mathcal{E}^2}{R} \left[t + \tau e^{-t/\tau} \right]_0^\tau = \left(\frac{\mathcal{E}^2}{R}\right) (\tau + \tau e^{-1} - \tau)$$

$$U_\mathcal{E} = \left(\frac{\mathcal{E}^2}{R}\right) \tau e^{-1} = \left(\frac{\mathcal{E}^2}{R}\right) \left(\frac{L}{R}\right) e^{-1} = \left(\frac{\mathcal{E}}{R}\right)^2 L e^{-1}$$

$$U_\mathcal{E} = \left(\frac{6.00 \text{ V}}{8.00 \Omega}\right)^2 (2.50 \text{ H})(0.3679) = 0.517 \text{ J}$$

$$\text{(c) } P_R = i^2 R = \left(\frac{\mathcal{E}}{R}\right)^2 (1 - e^{-t/\tau})^2 = \frac{\mathcal{E}^2}{R} (1 - 2e^{-t/\tau} + e^{-2t/\tau})$$

$$U_R = \int_0^\tau P_R dt = \frac{\mathcal{E}^2}{R} \int_0^\tau (1 - 2e^{-t/\tau} + e^{-2t/\tau}) dt = \frac{\mathcal{E}^2}{R} \left[t + 2\tau e^{-t/\tau} - \frac{\tau}{2} e^{-2t/\tau} \right]_0^\tau$$

$$U_R = \frac{\mathcal{E}^2}{R} \left[\tau + 2\tau e^{-1} - \frac{\tau}{2} e^{-2} - 2\tau + \frac{\tau}{2} \right] = \frac{\mathcal{E}^2}{R} \left[-\frac{\tau}{2} + 2\tau e^{-1} - \frac{\tau}{2} e^{-2} \right]$$

$$U_R = \left(\frac{\mathcal{E}^2}{2R}\right) \left(\frac{L}{R}\right) [-1 + 4e^{-1} - e^{-2}]$$

$$U_R = \left(\frac{\mathcal{E}}{R}\right)^2 \left(\frac{1}{2}L\right) [-1 + 4e^{-1} - e^{-2}] = \left(\frac{6.00 \text{ V}}{8.00 \Omega}\right)^2 \frac{1}{2}(2.50 \text{ H})(0.3362) = 0.236 \text{ J}$$

(d) EVALUATE: $U_\mathcal{E} = U_R + U_L$. (0.517 J = 0.236 J + 0.281 J)

The energy supplied by the battery equals the sum of the energy stored in the magnetic field of the inductor and the energy dissipated in the resistance of the inductor.

30.54. IDENTIFY: This is a decaying R - L circuit with $I_0 = \mathcal{E}/R$. $i(t) = I_0 e^{-(R/L)t}$.

SET UP: $\mathcal{E} = 60.0 \text{ V}$, $R = 240 \Omega$ and $L = 0.160 \text{ H}$. The rate at which energy stored in the inductor is decreasing is $iL di/dt$.

$$\text{EXECUTE: (a) } U = \frac{1}{2}LI_0^2 = \frac{1}{2}L \left(\frac{\mathcal{E}}{R}\right)^2 = \frac{1}{2}(0.160 \text{ H}) \left(\frac{60 \text{ V}}{240 \Omega}\right)^2 = 5.00 \times 10^{-3} \text{ J.}$$

$$(b) i = \frac{\mathcal{E}}{R} e^{-(R/L)t} \Rightarrow \frac{di}{dt} = -\frac{R}{L} i \Rightarrow \frac{dU_L}{dt} = iL \frac{di}{dt} = -Ri^2 = \frac{\mathcal{E}^2}{R} e^{-2(R/L)t} \cdot \frac{dU_L}{dt} = -\frac{(60 \text{ V})^2}{240 \Omega} e^{-2(240/0.160)(4.00 \times 10^{-4})} = -4.52 \text{ W}.$$

$$(c) \text{ In the resistor, } P_R = \frac{dU_R}{dt} = i^2 R = \frac{\mathcal{E}^2}{R} e^{-2(R/L)t} = \frac{(60 \text{ V})^2}{240 \Omega} e^{-2(240/0.160)(4.00 \times 10^{-4})} = 4.52 \text{ W}.$$

$$(d) P_R(t) = i^2 R = \frac{\mathcal{E}^2}{R} e^{-2(R/L)t}. U_R = \frac{\mathcal{E}^2}{R} \int_0^\infty e^{-2(R/L)t} dt = \frac{\mathcal{E}^2}{R} \frac{L}{2R} = \frac{(60 \text{ V})^2 (0.160 \text{ H})}{2(240 \Omega)^2} = 5.00 \times 10^{-3} \text{ J, which is the same as part (a).}$$

EVALUATE: During the decay of the current all the electrical energy originally stored in the inductor is dissipated in the resistor.

30.55. IDENTIFY and SET UP: Follow the procedure specified in the problem. $\frac{1}{2} Li^2$ is the energy stored in the inductor and $q^2/2C$ is the energy stored in the capacitor. The equation is $-iR - L \frac{di}{dt} - \frac{q}{C} = 0$.

EXECUTE: Multiplying by $-i$ gives $i^2 R + Li \frac{di}{dt} + \frac{qi}{C} = 0$. $\frac{d}{dt} U_L = \frac{d}{dt} \left(\frac{1}{2} Li^2 \right) = \frac{1}{2} L \frac{d}{dt} (i^2) = \frac{1}{2} L \left(2i \frac{di}{dt} \right) = Li \frac{di}{dt}$, the

second term. $\frac{d}{dt} U_C = \frac{d}{dt} \left(\frac{q^2}{2C} \right) = \frac{1}{2C} \frac{d}{dt} (q^2) = \frac{1}{2C} (2q) \frac{dq}{dt} = \frac{qi}{C}$, the third term. $i^2 R = P_R$, the rate at which

electrical energy is dissipated in the resistance. $\frac{d}{dt} U_L = P_L$, the rate at which the amount of energy stored in the

inductor is changing. $\frac{d}{dt} U_C = P_C$, the rate at which the amount of energy stored in the capacitor is changing.

EVALUATE: The equation says that $P_R + P_L + P_C = 0$; the net rate of change of energy in the circuit is zero. Note that at any given time one of P_C or P_L is negative. If the current and U_L are increasing the charge on the capacitor and U_C are decreasing, and vice versa.

30.56. IDENTIFY: The energy stored in a capacitor is $U_C = \frac{1}{2} C v^2$. The energy stored in an inductor is $U_L = \frac{1}{2} Li^2$. Energy conservation requires that the total stored energy be constant.

SET UP: The current is a maximum when the charge on the capacitor is zero and the energy stored in the capacitor is zero.

EXECUTE: (a) Initially $v = 16.0 \text{ V}$ and $i = 0$. $U_L = 0$ and $U_C = \frac{1}{2} C v^2 = \frac{1}{2} (5.00 \times 10^{-6} \text{ F})(16.0 \text{ V})^2 = 6.40 \times 10^{-4} \text{ J}$. The total energy stored is 0.640 mJ .

(b) The current is maximum when $q = 0$ and $U_C = 0$. $U_C + U_L = 6.40 \times 10^{-4} \text{ J}$ so $U_L = 6.40 \times 10^{-4} \text{ J}$.

$$\frac{1}{2} Li_{\max}^2 = 6.40 \times 10^{-4} \text{ J and } i_{\max} = \sqrt{\frac{2(6.40 \times 10^{-4} \text{ J})}{3.75 \times 10^{-3} \text{ H}}} = 0.584 \text{ A}.$$

EVALUATE: The maximum charge on the capacitor is $Q = CV = 80.0 \mu\text{C}$.

30.57. IDENTIFY and SET UP: Use $U_C = \frac{1}{2} CV_C^2$ (energy stored in a capacitor) to solve for C . Then use Eq.(30.22) and $\omega = 2\pi f$ to solve for the L that gives the desired current oscillation frequency.

EXECUTE: $V_C = 12.0 \text{ V}$; $U_C = \frac{1}{2} CV_C^2$ so $C = 2U_C/V_C^2 = 2(0.0160 \text{ J})/(12.0 \text{ V})^2 = 222 \mu\text{F}$

$$f = \frac{1}{2\pi\sqrt{LC}} \text{ so } L = \frac{1}{(2\pi f)^2 C}$$

$$f = 3500 \text{ Hz gives } L = 9.31 \mu\text{H}$$

EVALUATE: f is in Hz and ω is in rad/s; we must be careful not to confuse the two.

30.58. IDENTIFY: Apply energy conservation to the circuit.

SET UP: For a capacitor $V = q/C$ and $U = q^2/2C$. For an inductor $U = \frac{1}{2} Li^2$

EXECUTE: (a) $V_{\max} = \frac{Q}{C} = \frac{6.00 \times 10^{-6} \text{ C}}{2.50 \times 10^{-4} \text{ F}} = 0.0240 \text{ V}$.

(b) $\frac{1}{2} Li_{\max}^2 = \frac{Q^2}{2C}$, so $i_{\max} = \frac{Q}{\sqrt{LC}} = \frac{6.00 \times 10^{-6} \text{ C}}{\sqrt{(0.0600 \text{ H})(2.50 \times 10^{-4} \text{ F})}} = 1.55 \times 10^{-3} \text{ A}$

(c) $U_{\max} = \frac{1}{2} Li_{\max}^2 = \frac{1}{2} (0.0600 \text{ H})(1.55 \times 10^{-3} \text{ A})^2 = 7.21 \times 10^{-8} \text{ J}$.

(d) If $i = \frac{1}{2}i_{\max}$ then $U_L = \frac{1}{4}U_{\max} = 1.80 \times 10^{-8} \text{ J}$ and $U_C = \frac{3}{4}U_{\max} = \frac{(\sqrt{(3/4)Q})^2}{2C} = \frac{q^2}{2C}$. This gives $q = \sqrt{\frac{3}{4}}Q = 5.20 \times 10^{-6} \text{ C}$.

EVALUATE: $U_{\max} = \frac{1}{2}Li^2 + \frac{1}{2}\frac{q^2}{C}$ for all times.

30.59. IDENTIFY: Set $U_B = K$, where $K = \frac{1}{2}mv^2$.

SET UP: The energy density in the magnetic field is $u_B = B^2/2\mu_0$. Consider volume $V = 1 \text{ m}^3$ of sunspot material.

EXECUTE: The energy density in the sunspot is $u_B = B^2/2\mu_0 = 6.366 \times 10^4 \text{ J/m}^3$. The total energy stored in volume V of the sunspot is $U_B = u_B V$. The mass of the material in volume V of the sunspot is $m = \rho V$.

$K = U_B$ so $\frac{1}{2}mv^2 = U_B$. $\frac{1}{2}\rho V v^2 = u_B V$. The volume divides out, and $v = \sqrt{2u_B/\rho} = 2 \times 10^4 \text{ m/s}$.

EVALUATE: The speed we calculated is about 30 times smaller than the escape speed.

30.60. IDENTIFY: $i(t)$ is given by Eq.(30.14).

SET UP: The graph shows $V = 0$ at $t = 0$ and V approaches the constant value of 25 V at large times.

EXECUTE: (a) The voltage behaves the same as the current. Since V_R is proportional to i , the scope must be across the 150 Ω resistor.

(b) From the graph, as $t \rightarrow \infty$, $V_R \rightarrow 25 \text{ V}$, so there is no voltage drop across the inductor, so its internal resistance must be zero. $V_R = V_{\max}(1 - e^{-t/\tau})$. When $t = \tau$, $V_R = V_{\max}\left(1 - \frac{1}{e}\right) \approx 0.63V_{\max}$. From the graph, $V = 0.63V_{\max} = 16 \text{ V}$ at $t \approx 0.5 \text{ ms}$. Therefore $\tau = 0.5 \text{ ms}$. $L/R = 0.5 \text{ ms}$ gives $L = (0.5 \text{ ms})(150 \Omega) = 0.075 \text{ H}$.

(c) The graph if the scope is across the inductor is sketched in Figure 30.60.

EVALUATE: At all times $V_R + V_L = 25.0 \text{ V}$. At $t = 0$ all the battery voltage appears across the inductor since $i = 0$. At $t \rightarrow \infty$ all the battery voltage is across the resistance, since $di/dt = 0$.

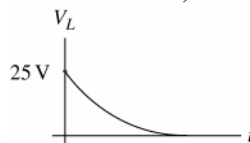


Figure 30.60

30.61. IDENTIFY and SET UP: The current grows in the circuit as given by Eq.(30.14). In an R - L circuit the full emf initially is across the inductance and after a long time is totally across the resistance. A solenoid in a circuit is represented as a resistance in series with an inductance. Apply the loop rule to the circuit; the voltage across a resistance is given by Ohm's law.

EXECUTE: (a) In the R - L circuit the voltage across the resistor starts at zero and increases to the battery voltage. The voltage across the solenoid (inductor) starts at the battery voltage and decreases to zero. In the graph, the voltage drops, so the oscilloscope is across the solenoid.

(b) At $t \rightarrow \infty$ the current in the circuit approaches its final, constant value. The voltage doesn't go to zero because the solenoid has some resistance R_L . The final voltage across the solenoid is IR_L , where I is the final current in the circuit.

(c) The emf of the battery is the initial voltage across the inductor, 50 V. Just after the switch is closed, the current is zero and there is no voltage drop across any of the resistance in the circuit.

(d) As $t \rightarrow \infty$, $\mathcal{E} - IR - IR_L = 0$

$\mathcal{E} = 50 \text{ V}$ and from the graph $IR_L = 15 \text{ V}$ (the final voltage across the inductor), so $IR = 35 \text{ V}$ and $I = (35 \text{ V})/R = 3.5 \text{ A}$

(e) $IR_L = 15 \text{ V}$, so $R_L = (15 \text{ V})/(3.5 \text{ A}) = 4.3 \Omega$

$\mathcal{E} - V_L - iR = 0$, where V_L includes the voltage across the resistance of the solenoid.

$V_L = \mathcal{E} - iR$, $i = \frac{\mathcal{E}}{R_{\text{tot}}}(1 - e^{-t/\tau})$, so $V_L = \mathcal{E}\left[1 - \frac{R}{R_{\text{tot}}}(1 - e^{-t/\tau})\right]$

$\mathcal{E} = 50 \text{ V}$, $R = 10 \Omega$, $R_{\text{tot}} = 14.3 \Omega$, so when $t = \tau$, $V_L = 27.9 \text{ V}$. From the graph, V_L has this value when $t = 3.0 \text{ ms}$ (read approximately from the graph), so $\tau = L/R_{\text{tot}} = 3.0 \text{ ms}$. Then $L = (3.0 \text{ ms})(14.3 \Omega) = 43 \text{ mH}$.

EVALUATE: At $t = 0$ there is no current and the 50 V measured by the oscilloscope is the induced emf due to the inductance of the solenoid. As the current grows, there are voltage drops across the two resistances in the circuit.

We derived an equation for V_L , the voltage across the solenoid. At $t = 0$ it gives $V_L = \mathcal{E}$ and at $t \rightarrow \infty$ it gives $V_L = \mathcal{E}R/R_{\text{tot}} = iR$.

30.62. IDENTIFY: At $t = 0$, $i = 0$ through each inductor. At $t \rightarrow \infty$, the voltage is zero across each inductor.

SET UP: In each case redraw the circuit. At $t = 0$ replace each inductor by a break in the circuit and at $t \rightarrow \infty$ replace each inductor by a wire.

EXECUTE: (a) Initially the inductor blocks current through it, so the simplified equivalent circuit is shown in Figure 30.62a. $i = \frac{\mathcal{E}}{R} = \frac{50 \text{ V}}{150 \Omega} = 0.333 \text{ A}$. $V_1 = (100 \Omega)(0.333 \text{ A}) = 33.3 \text{ V}$. $V_4 = (50 \Omega)(0.333 \text{ A}) = 16.7 \text{ V}$. $V_3 = 0$

since no current flows through it. $V_2 = V_4 = 16.7 \text{ V}$, since the inductor is in parallel with the 50Ω resistor.

$A_1 = A_3 = 0.333 \text{ A}$, $A_2 = 0$.

(b) Long after S is closed, steady state is reached, so the inductor has no potential drop across it. The simplified circuit is sketched in Figure 30.62b. $i = \mathcal{E}/R = \frac{50 \text{ V}}{130 \Omega} = 0.385 \text{ A}$. $V_1 = (100 \Omega)(0.385 \text{ A}) = 38.5 \text{ V}$; $V_2 = 0$;

$V_3 = V_4 = 50 \text{ V} - 38.5 \text{ V} = 11.5 \text{ V}$. $i_1 = 0.385 \text{ A}$; $i_2 = \frac{11.5 \text{ V}}{75 \Omega} = 0.153 \text{ A}$; $i_3 = \frac{11.5 \text{ V}}{50 \Omega} = 0.230 \text{ A}$.

EVALUATE: Just after the switch is closed the current through the battery is 0.333 A . After a long time the current through the battery is 0.385 A . After a long time there is an additional current path, the equivalent resistance of the circuit is decreased and the current has increased.

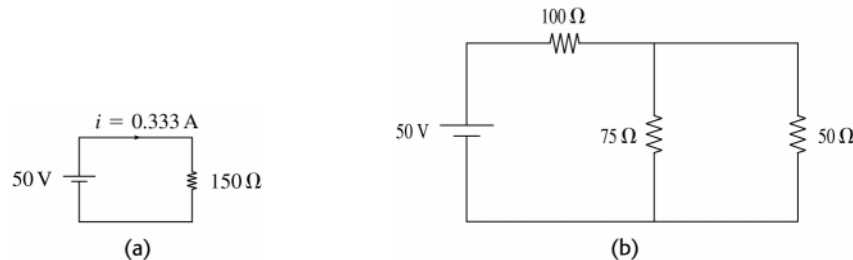


Figure 30.62

30.63. IDENTIFY and SET UP: Just after the switch is closed, the current in each branch containing an inductor is zero and the voltage across any capacitor is zero. The inductors can be treated as breaks in the circuit and the capacitors can be replaced by wires. After a long time there is no voltage across each inductor and no current in any branch containing a capacitor. The inductors can be replaced by wires and the capacitors by breaks in the circuit.

EXECUTE: (a) Just after the switch is closed the voltage V_5 across the capacitor is zero and there is also no current through the inductor, so $V_3 = 0$. $V_2 + V_3 = V_4 = V_5$, and since $V_5 = 0$ and $V_3 = 0$, V_4 and V_2 are also zero.

$V_4 = 0$ means V_3 reads zero. V_1 then must equal 40.0 V , and this means the current read by A_1 is $(40.0 \text{ V})/(50.0 \Omega) = 0.800 \text{ A}$. $A_2 + A_3 + A_4 = A_1$, but $A_2 = A_3 = 0$ so $A_4 = A_1 = 0.800 \text{ A}$. $A_1 = A_4 = 0.800 \text{ A}$; all other ammeters read zero. $V_1 = 40.0 \text{ V}$ and all other voltmeters read zero.

(b) After a long time the capacitor is fully charged so $A_4 = 0$. The current through the inductor isn't changing, so $V_2 = 0$. The currents can be calculated from the equivalent circuit that replaces the inductor by a short circuit, as shown in Figure 30.63a.

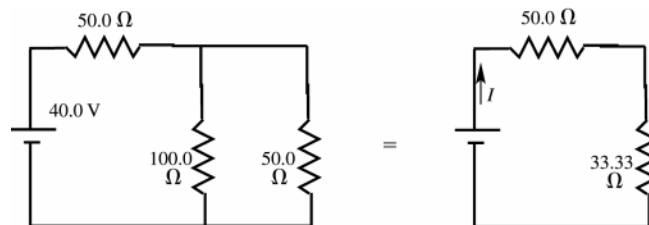


Figure 30.63a

$I = (40.0 \text{ V})/(83.33 \Omega) = 0.480 \text{ A}$; A_1 reads 0.480 A

$V_1 = I(50.0 \Omega) = 24.0 \text{ V}$

The voltage across each parallel branch is $40.0 \text{ V} - 24.0 \text{ V} = 16.0 \text{ V}$

$V_2 = 0$, $V_3 = V_4 = V_5 = 16.0 \text{ V}$

$V_3 = 16.0 \text{ V}$ means A_2 reads 0.160 A . $V_4 = 16.0 \text{ V}$ means A_3 reads 0.320 A . A_4 reads zero. Note that $A_2 + A_3 = A_1$.

(c) $V_5 = 16.0 \text{ V}$ so $Q = CV = (12.0 \mu\text{F})(16.0 \text{ V}) = 192 \mu\text{C}$

(d) At $t = 0$ and $t \rightarrow \infty$, $V_2 = 0$. As the current in this branch increases from zero to 0.160 A the voltage V_2 reflects the rate of change of the current. The graph is sketched in Figure 30.63b.

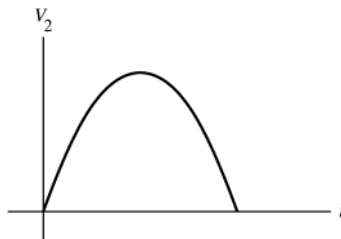


Figure 30.63b

EVALUATE: This reduction of the circuit to resistor networks only apply at $t = 0$ and $t \rightarrow \infty$. At intermediate times the analysis is complicated.

30.64. IDENTIFY: At all times $v_1 + v_2 = 25.0 \text{ V}$. The voltage across the resistor depends on the current through it and the voltage across the inductor depends on the rate at which the current through it is changing.

SET UP: Immediately after closing the switch the current through the inductor is zero. After a long time the current is no longer changing.

EXECUTE: (a) $i = 0$ so $v_1 = 0$ and $v_2 = 25.0 \text{ V}$. The ammeter reading is $A = 0$.

(b) After a long time, $v_2 = 0$ and $v_1 = 25.0 \text{ V}$. $v_1 = iR$ and $i = \frac{v_1}{R} = \frac{25.0 \text{ V}}{15.0 \Omega} = 1.67 \text{ A}$. The ammeter reading is $A = 1.67 \text{ A}$.

(c) None of the answers in (a) and (b) depend on L so none of them would change.

EVALUATE: The inductance L of the circuit affects the rate at which current reaches its final value. But after a long time the inductor doesn't affect the circuit and the final current does not depend on L .

30.65. IDENTIFY: At $t = 0$, $i = 0$ through each inductor. At $t \rightarrow \infty$, the voltage is zero across each inductor.

SET UP: In each case redraw the circuit. At $t = 0$ replace each inductor by a break in the circuit and at $t \rightarrow \infty$ replace each inductor by a wire.

EXECUTE: (a) Just after the switch is closed there is no current through either inductor and they act like breaks in the circuit. The current is the same through the 40.0Ω and 15.0Ω resistors and is equal to

$$(25.0 \text{ V}) / (40.0 \Omega + 15.0 \Omega) = 0.455 \text{ A}. A_1 = A_4 = 0.455 \text{ A}; A_2 = A_3 = 0.$$

(b) After a long time the currents are constant, there is no voltage across either inductor, and each inductor can be treated as a short-circuit. The circuit is equivalent to the circuit sketched in Figure 30.65.

$$I = (25.0 \text{ V}) / (42.73 \Omega) = 0.585 \text{ A}. A_1 \text{ reads } 0.585 \text{ A}. \text{ The voltage across each parallel branch is}$$

$$25.0 \text{ V} - (0.585 \text{ A})(40.0 \Omega) = 1.60 \text{ V}. A_2 \text{ reads } (1.60 \text{ V}) / (5.0 \Omega) = 0.320 \text{ A}. A_3 \text{ reads}$$

$$(1.60 \text{ V}) / (10.0 \Omega) = 0.160 \text{ A}. A_4 \text{ reads } (1.60 \text{ V}) / (15.0 \Omega) = 0.107 \text{ A}.$$

EVALUATE: Just after the switch is closed the current through the battery is 0.455 A. After a long time the current through the battery is 0.585 A. After a long time there are additional current paths, the equivalent resistance of the circuit is decreased and the current has increased.

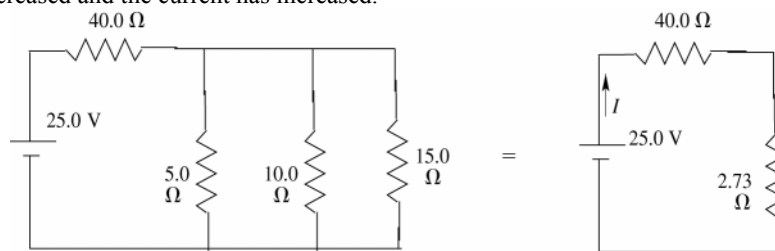


Figure 30.65

30.66. IDENTIFY: Closing S_2 and simultaneously opening S_1 produces an L - C circuit with initial current through the inductor of 3.50 A. When the current is a maximum the charge q on the capacitor is zero and when the charge q is a maximum the current is zero. Conservation of energy says that the maximum energy $\frac{1}{2}Li_{\text{max}}^2$ stored in the inductor equals the maximum energy $\frac{1}{2}\frac{q_{\text{max}}^2}{C}$ stored in the capacitor.

SET UP: $i_{\text{max}} = 3.50 \text{ A}$, the current in the inductor just after the switch is closed.

EXECUTE: (a) $\frac{1}{2}Li_{\max}^2 = \frac{1}{2}\frac{q_{\max}^2}{C}$.

$$q_{\max} = (\sqrt{LC})i_{\max} = \sqrt{(2.0 \times 10^{-3} \text{ H})(5.0 \times 10^{-6} \text{ F})(3.50 \text{ A})} = 3.50 \times 10^{-4} \text{ C} = 0.350 \text{ mC}.$$

(b) When q is maximum, $i = 0$.

EVALUATE: In the final circuit the current will oscillate.

- 30.67. IDENTIFY:** Apply the loop rule to each parallel branch. The voltage across a resistor is given by iR and the voltage across an inductor is given by $L|di/dt|$. The rate of change of current through the inductor is limited.

SET UP: With S closed the circuit is sketched in Figure 30.67a.

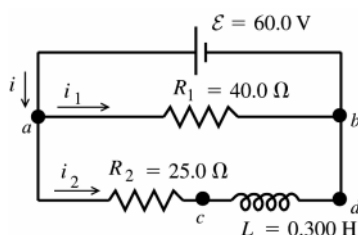


Figure 30.67a

The rate of change of the current through the inductor is limited by the induced emf. Just after the switch is closed the current in the inductor has not had time to increase from zero, so $i_2 = 0$.

EXECUTE: (a) $\mathcal{E} - v_{ab} = 0$, so $v_{ab} = 60.0 \text{ V}$

(b) The voltage drops across R_1 , as we travel through the resistor in the direction of the current, so point a is at higher potential.

(c) $i_2 = 0$ so $v_{R_2} = i_2 R_2 = 0$

$\mathcal{E} - v_{R_2} - v_L = 0$ so $v_L = \mathcal{E} = 60.0 \text{ V}$

(d) The voltage rises when we go from b to a through the emf, so it must drop when we go from a to b through the inductor. Point c must be at higher potential than point d .

(e) After the switch has been closed a long time, $\frac{di_2}{dt} \rightarrow 0$ so $v_L = 0$. Then $\mathcal{E} - v_{R_2} = 0$ and $i_2 R_2 = \mathcal{E}$

$$\text{so } i_2 = \frac{\mathcal{E}}{R_2} = \frac{60.0 \text{ V}}{25.0 \Omega} = 2.40 \text{ A}.$$

SET UP: The rate of change of the current through the inductor is limited by the induced emf. Just after the switch is opened again the current through the inductor hasn't had time to change and is still $i_2 = 2.40 \text{ A}$. The circuit is sketched in Figure 30.67b.

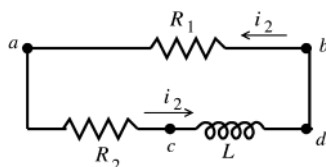


Figure 30.67b

EXECUTE: The current through R_1 is $i_2 = 2.40 \text{ A}$, in the direction b to a . Thus

$$v_{ab} = -i_2 R_1 = -(2.40 \text{ A})(40.0 \Omega)$$

$$v_{ab} = -96.0 \text{ V}$$

(f) Point where current enters resistor is at higher potential; point b is at higher potential.

(g) $v_L - v_{R_1} - v_{R_2} = 0$

$$v_L = v_{R_1} + v_{R_2}$$

$$v_{R_1} = -v_{ab} = 96.0 \text{ V}; v_{R_2} = i_2 R_2 = (2.40 \text{ A})(25.0 \Omega) = 60.0 \text{ V}$$

$$\text{Then } v_L = v_{R_1} + v_{R_2} = 96.0 \text{ V} + 60.0 \text{ V} = 156 \text{ V}.$$

As you travel counterclockwise around the circuit in the direction of the current, the voltage drops across each resistor, so it must rise across the inductor and point d is at higher potential than point c . The current is decreasing, so the induced emf in the inductor is directed in the direction of the current. Thus, $v_{cd} = -156 \text{ V}$.

(h) Point d is at higher potential.

EVALUATE: The voltage across R_1 is constant once the switch is closed. In the branch containing R_2 , just after S is closed the voltage drop is all across L and after a long time it is all across R_2 . Just after S is opened the same current flows in the single loop as had been flowing through the inductor and the sum of the voltage across the resistors equals the voltage across the inductor. This voltage dies away, as the energy stored in the inductor is dissipated in the resistors.

30.68. IDENTIFY: Apply the loop rule to the two loops. The current through the inductor doesn't change abruptly.

SET UP: For the inductor $|\mathcal{E}| = L \left| \frac{di}{dt} \right|$ and \mathcal{E} is directed to oppose the change in current.

EXECUTE: (a) Switch is closed, then at some later time

$$\frac{di}{dt} = 50.0 \text{ A/s} \Rightarrow v_{cd} = L \frac{di}{dt} = (0.300 \text{ H})(50.0 \text{ A/s}) = 15.0 \text{ V}.$$

The top circuit loop: $60.0 \text{ V} = i_1 R_1 \Rightarrow i_1 = \frac{60.0 \text{ V}}{40.0 \Omega} = 1.50 \text{ A}.$

The bottom loop: $60 \text{ V} - i_2 R_2 - 15.0 \text{ V} = 0 \Rightarrow i_2 = \frac{45.0 \text{ V}}{25.0 \Omega} = 1.80 \text{ A}.$

(b) After a long time: $i_2 = \frac{60.0 \text{ V}}{25.0 \Omega} = 2.40 \text{ A}$, and immediately when the switch is opened, the inductor maintains this current, so $i_1 = i_2 = 2.40 \text{ A}.$

EVALUATE: The current through R_1 changes abruptly when the switch is closed.

30.69. IDENTIFY and SET UP: The circuit is sketched in Figure 30.69a. Apply the loop rule. Just after S_1 is closed, $i = 0$. After a long time i has reached its final value and $di/dt = 0$. The voltage across a resistor depends on i and the voltage across an inductor depends on di/dt .

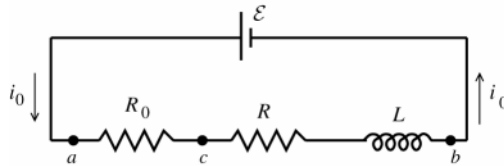


Figure 30.69a

EXECUTE: (a) At time $t = 0$, $i_0 = 0$ so $v_{ac} = i_0 R_0 = 0$. By the loop rule $\mathcal{E} - v_{ac} - v_{cb} = 0$, so $v_{cb} = \mathcal{E} - v_{ac} = \mathcal{E} = 36.0 \text{ V}$. ($i_0 R = 0$ so this potential difference of 36.0 V is across the inductor and is an induced emf produced by the changing current.)

(b) After a long time $\frac{di_0}{dt} \rightarrow 0$ so the potential $-L \frac{di_0}{dt}$ across the inductor becomes zero. The loop rule gives

$$\mathcal{E} - i_0(R_0 + R) = 0.$$

$$i_0 = \frac{\mathcal{E}}{R_0 + R} = \frac{36.0 \text{ V}}{50.0 \Omega + 150 \Omega} = 0.180 \text{ A}$$

$$v_{ac} = i_0 R_0 = (0.180 \text{ A})(50.0 \Omega) = 9.0 \text{ V}$$

Thus $v_{cb} = i_0 R + L \frac{di_0}{dt} = (0.180 \text{ A})(150 \Omega) + 0 = 27.0 \text{ V}$ (Note that $v_{ac} + v_{cb} = \mathcal{E}$.)

(c) $\mathcal{E} - v_{ac} - v_{cb} = 0$

$$\mathcal{E} - iR_0 - iR - L \frac{di}{dt} = 0$$

$$L \frac{di}{dt} = \mathcal{E} - i(R_0 + R) \text{ and } \left(\frac{L}{R + R_0} \right) \frac{di}{dt} = -i + \frac{\mathcal{E}}{R + R_0}$$

$$\frac{di}{-i + \mathcal{E}/(R + R_0)} = \left(\frac{R + R_0}{L} \right) dt$$

Integrate from $t = 0$, when $i = 0$, to t , when $i = i_0$: $\int_0^{i_0} \frac{di}{-i + \mathcal{E}/(R + R_0)} = \frac{R + R_0}{L} \int_0^t dt = -\ln \left[-i + \frac{\mathcal{E}}{R + R_0} \right]_0^{i_0} = \left(\frac{R + R_0}{L} \right) t$,

$$\text{so } \ln \left(-i_0 + \frac{\mathcal{E}}{R + R_0} \right) - \ln \left(\frac{\mathcal{E}}{R + R_0} \right) = - \left(\frac{R + R_0}{L} \right) t$$

$$\ln \left(\frac{-i_0 + \mathcal{E}/(R + R_0)}{\mathcal{E}/(R + R_0)} \right) = - \left(\frac{R + R_0}{L} \right) t$$

Taking exponentials of both sides gives $\frac{-i_0 + \mathcal{E}/(R + R_0)}{\mathcal{E}/(R + R_0)} = e^{-(R + R_0)t/L}$ and $i_0 = \frac{\mathcal{E}}{R + R_0} (1 - e^{-(R + R_0)t/L})$

Substituting in the numerical values gives $i_0 = \frac{36.0 \text{ V}}{50 \Omega + 150 \Omega} (1 - e^{-(200 \Omega/4.00 \text{ H})t}) = (0.180 \text{ A})(1 - e^{-t/0.020 \text{ s}})$

At $t \rightarrow 0$, $i_0 = (0.180 \text{ A})(1-1) = 0$ (agrees with part (a)). At $t \rightarrow \infty$, $i_0 = (0.180 \text{ A})(1-0) = 0.180 \text{ A}$ (agrees with part (b)).

$$v_{ac} = i_0 R_0 = \frac{\mathcal{E} R_0}{R + R_0} (1 - e^{-(R+R_0)t/L}) = 9.0 \text{ V} (1 - e^{-t/0.020 \text{ s}})$$

$$v_{cb} = \mathcal{E} - v_{ac} = 36.0 \text{ V} - 9.0 \text{ V} (1 - e^{-t/0.020 \text{ s}}) = 9.0 \text{ V} (3.00 + e^{-t/0.020 \text{ s}})$$

At $t \rightarrow 0$, $v_{ac} = 0$, $v_{cb} = 36.0 \text{ V}$ (agrees with part (a)). At $t \rightarrow \infty$, $v_{ac} = 9.0 \text{ V}$, $v_{cb} = 27.0 \text{ V}$ (agrees with part (b)).

The graphs are given in Figure 30.69b.

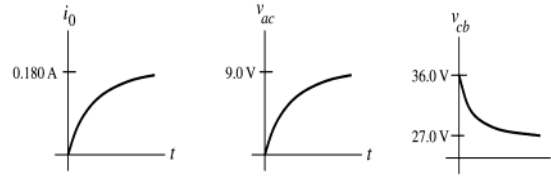


Figure 30.69b

EVALUATE: The expression for $i(t)$ we derived becomes Eq.(30.14) if the two resistors R_0 and R in series are replaced by a single equivalent resistance $R_0 + R$.

30.70. IDENTIFY: Apply the loop rule. The current through the inductor doesn't change abruptly.

SET UP: With S_2 closed, v_{cb} must be zero.

EXECUTE: (a) Immediately after S_2 is closed, the inductor maintains the current $i = 0.180 \text{ A}$ through R . The loop rule around the outside of the circuit yields

$$\mathcal{E} + \mathcal{E}_L - iR - i_0 R_0 = 36.0 \text{ V} + (0.18 \text{ A})(150 \Omega) - (0.18 \text{ A})(150 \Omega) - i_0 (50 \Omega) = 0. \quad i_0 = \frac{36 \text{ V}}{50 \Omega} = 0.720 \text{ A}.$$

$$v_{ac} = (0.72 \text{ A})(50 \text{ V}) = 36.0 \text{ V} \text{ and } v_{cb} = 0.$$

(b) After a long time, $v_{ac} = 36.0 \text{ V}$, and $v_{cb} = 0$. Thus $i_0 = \frac{\mathcal{E}}{R_0} = \frac{36.0 \text{ V}}{50 \Omega} = 0.720 \text{ A}$, $i_R = 0$ and $i_{s2} = 0.720 \text{ A}$.

(c) $i_0 = 0.720 \text{ A}$, $i_R(t) = \frac{\mathcal{E}}{R_{\text{total}}} e^{-(R/L)t}$ and $i_R(t) = (0.180 \text{ A}) e^{-(12.5 \text{ s}^{-1})t}$.

$i_{s2}(t) = (0.720 \text{ A}) - (0.180 \text{ A}) e^{-(12.5 \text{ s}^{-1})t} = (0.180 \text{ A}) (4 - e^{-(12.5 \text{ s}^{-1})t})$. The graphs of the currents are given in Figure 30.70.

EVALUATE: R_0 is in a loop that contains just \mathcal{E} and R_0 , so the current through R_0 is constant. After a long time the current through the inductor isn't changing and the voltage across the inductor is zero. Since v_{cb} is zero, the voltage across R must be zero and i_R becomes zero.

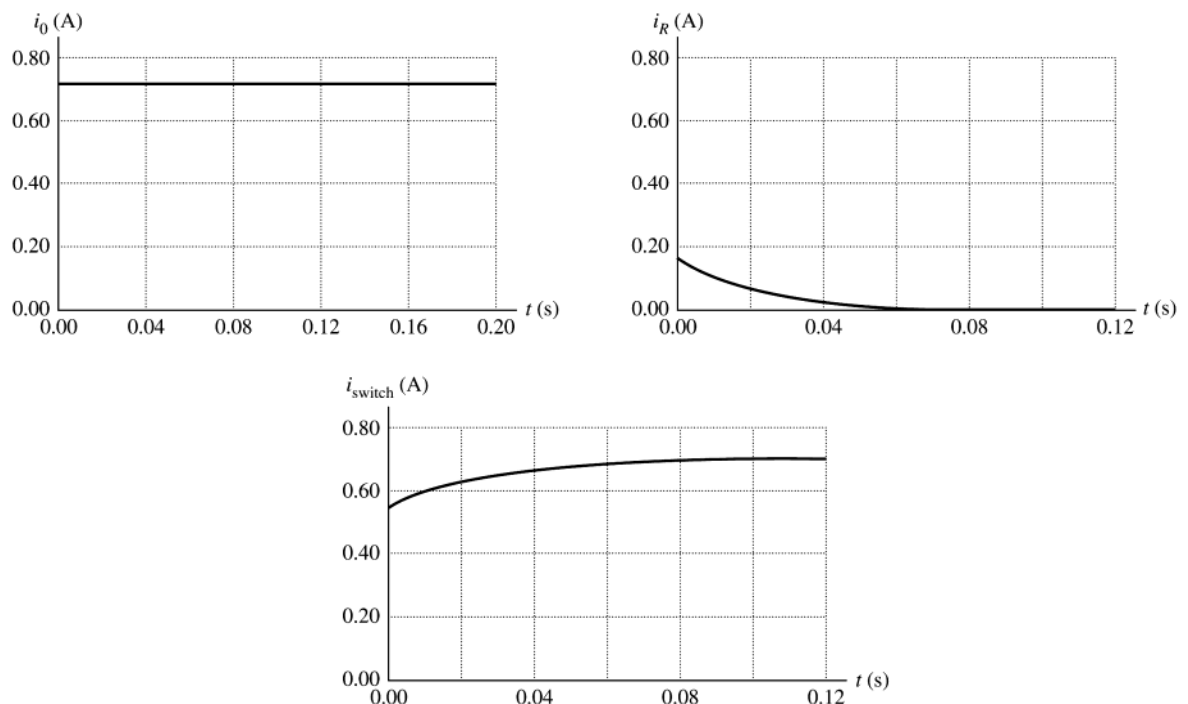


Figure 30.70

30.71. IDENTIFY: The current through an inductor doesn't change abruptly. After a long time the current isn't changing and the voltage across each inductor is zero.

SET UP: Problem 30.47 shows how to find the equivalent inductance of inductors in series and parallel.

EXECUTE: (a) Just after the switch is closed there is no current in the inductors. There is no current in the resistors so there is no voltage drop across either resistor. A reads zero and V reads 20.0 V.

(b) After a long time the currents are no longer changing, there is no voltage across the inductors, and the inductors can be replaced by short-circuits. The circuit becomes equivalent to the circuit shown in Figure 30.71a.

$I = (20.0 \text{ V}) / (75.0 \Omega) = 0.267 \text{ A}$. The voltage between points a and b is zero, so the voltmeter reads zero.

(c) Use the results of Problem 30.49 to combine the inductor network into its equivalent, as shown in Figure 30.71b.

$R = 75.0 \Omega$ is the equivalent resistance. Eq.(30.14) says $i = (\mathcal{E}/R)(1 - e^{-t/\tau})$ with

$\tau = L/R = (10.8 \text{ mH}) / (75.0 \Omega) = 0.144 \text{ ms}$. $\mathcal{E} = 20.0 \text{ V}$, $R = 75.0 \Omega$, $t = 0.115 \text{ ms}$ so $i = 0.147 \text{ A}$.

$V_R = iR = (0.147 \text{ A})(75.0 \Omega) = 11.0 \text{ V}$. $20.0 \text{ V} - V_R - V_L = 0$ and $V_L = 20.0 \text{ V} - V_R = 9.0 \text{ V}$. The ammeter reads 0.147 A and the voltmeter reads 9.0 V.

EVALUATE: The current through the battery increases from zero to a final value of 0.267 A. The voltage across the inductor network drops from 20.0 V to zero.

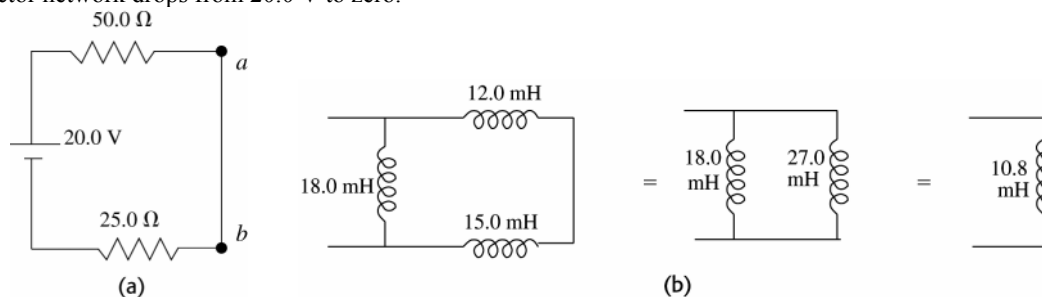


Figure 30.71

30.72. IDENTIFY: At steady state with the switch in position 1, no current flows to the capacitors and the inductors can be replaced by wires. Apply conservation of energy to the circuit with the switch in position 2.

SET UP: Replace the series combinations of inductors and capacitors by their equivalents. For the inductors use the results of Problem 30.47.

EXECUTE: (a) At steady state $i = \frac{\mathcal{E}}{R} = \frac{75.0 \text{ V}}{125 \Omega} = 0.600 \text{ A}$.

(b) The equivalent circuit capacitance of the two capacitors is given by $\frac{1}{C_s} = \frac{1}{25 \mu\text{F}} + \frac{1}{35 \mu\text{F}}$ and $C_s = 14.6 \mu\text{F}$.

$L_s = 15.0 \text{ mH} + 5.0 \text{ mH} = 20.0 \text{ mH}$. The equivalent circuit is sketched in Figure 30.72a.

Energy conservation: $\frac{q^2}{2C} = \frac{1}{2}Li_0^2$. $q = i_0\sqrt{LC} = (0.600 \text{ A})\sqrt{(20 \times 10^{-3} \text{ H})(14.6 \times 10^{-6} \text{ F})} = 3.24 \times 10^{-4} \text{ C}$. As shown in Figure 30.72b, the capacitors have their maximum charge at $t = T/4$.

$t = \frac{1}{4}T = \frac{1}{4}(2\pi\sqrt{LC}) = \frac{\pi}{2}\sqrt{LC} = \frac{\pi}{2}\sqrt{(20 \times 10^{-3} \text{ H})(14.6 \times 10^{-6} \text{ F})} = 8.49 \times 10^{-4} \text{ s}$

EVALUATE: With the switch closed the battery stores energy in the inductors. This then is the energy in the L - C circuit when the switch is in position 2.

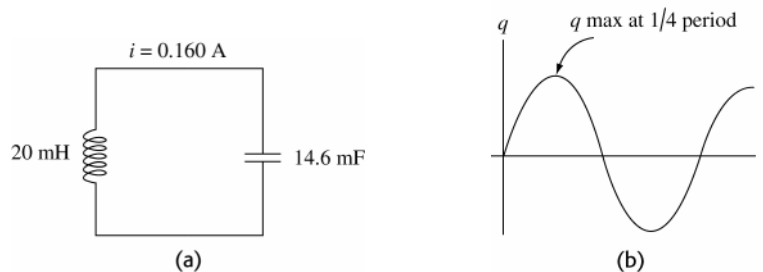


Figure 30.72

30.73. IDENTIFY: Follow the steps specified in the problem.

SET UP: Find the flux through a ring of height h , radius r and thickness dr . Example 28.19 shows that $B = \frac{\mu_0 Ni}{2\pi r}$ inside the toroid.

EXECUTE: (a) $\Phi_B = \int_a^b B(h dr) = \int_a^b \left(\frac{\mu_0 Ni}{2\pi r} \right) (h dr) = \frac{\mu_0 Nih}{2\pi} \int_a^b \frac{dr}{r} = \frac{\mu_0 Nih}{2\pi} \ln(b/a).$

(b) $L = \frac{N\Phi_B}{i} = \frac{\mu_0 N^2 h}{2\pi} \ln(b/a).$

(c) $\ln(b/a) = \ln(1 - (b-a)/a) \approx \frac{b-a}{a} + \frac{(b-a)^2}{2a^2} + \dots \Rightarrow L \approx \frac{\mu_0 N^2 h}{2\pi} \left(\frac{b-a}{a} \right).$

EVALUATE: $h(b-a)$ is the cross-sectional area A of the toroid and a is approximately the radius r , so this result is approximately the same as the result derived in Example 30.3.

30.74. IDENTIFY: The direction of the current induced in circuit A is given by Lenz's law.

SET UP: When the switch is closed current flows counterclockwise in the circuit on the left, from the positive plate of the capacitor. The current decreases as a function of time, as the charge and voltage of the capacitor decrease.

EXECUTE: At loop A the magnetic field from the wire of the other circuit adjacent to A is into the page. The magnetic field of this current is decreasing, as the current decreases. Therefore, the magnetic field of the induced current in A is directed into the page inside A and to produce a magnetic field in this direction the induced current is clockwise.

EVALUATE: The magnitude of the emf induced in circuit A decreases with time after the switch is closed, because the rate of change of the current in the other circuit decreases.

30.75. (a) IDENTIFY and SET UP: With switch S closed the circuit is shown in Figure 30.75a.

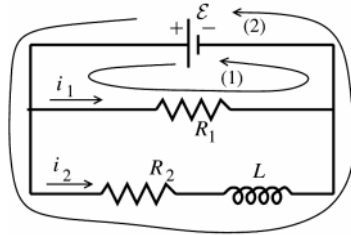


Figure 30.75a

Apply the loop rule to loops 1 and 2.

EXECUTE:

loop 1

$$\mathcal{E} - i_1 R_1 = 0$$

$$i_1 = \frac{\mathcal{E}}{R_1} \text{ (independent of } t\text{)}$$

loop (2)

$$\mathcal{E} - i_2 R_2 - L \frac{di_2}{dt} = 0$$

This is in the form of equation (30.12), so the solution is analogous to Eq.(30.14): $i_2 = \frac{\mathcal{E}}{R_2} (1 - e^{-R_2 t/L})$

(b) EVALUATE: The expressions derived in part (a) give that as $t \rightarrow \infty$, $i_1 = \frac{\mathcal{E}}{R_1}$ and $i_2 = \frac{\mathcal{E}}{R_2}$. Since $\frac{di_2}{dt} \rightarrow 0$ at steady-state, the inductance then has no effect on the circuit. The current in R_1 is constant; the current in R_2 starts at zero and rises to \mathcal{E}/R_2 .

(c) IDENTIFY and SET UP: The circuit now is as shown in Figure 30.75b.

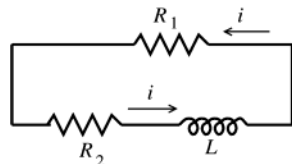


Figure 30.75b

Let $t = 0$ now be when S is opened.

$$\text{At } t = 0, i = \frac{\mathcal{E}}{R_2}.$$

Apply the loop rule to the single current loop.

EXECUTE: $-i(R_1 + R_2) - L \frac{di}{dt} = 0.$ (Now $\frac{di}{dt}$ is negative.)

$$L \frac{di}{dt} = -i(R_1 + R_2) \text{ gives } \frac{di}{i} = -\left(\frac{R_1 + R_2}{L} \right) dt$$

Integrate from $t = 0$, when $i = I_0 = \mathcal{E}/R_2$, to t .

$$\int_{I_0}^i \frac{di}{i} = -\left(\frac{R_1 + R_2}{L} \right) \int_0^t dt \text{ and } \ln\left(\frac{i}{I_0} \right) = -\left(\frac{R_1 + R_2}{L} \right) t$$

Taking exponentials of both sides of this equation gives $i = I_0 e^{-(R_1 + R_2)t/L} = \frac{\mathcal{E}}{R_2} e^{-(R_1 + R_2)t/L}$

(d) IDENTIFY and SET UP: Use the equation derived in part (c) and solve for R_2 and \mathcal{E} .

EXECUTE: $L = 22.0 \text{ H}$

$$R_{R_1} = \frac{V^2}{R_1} = 40.0 \text{ W gives } R_1 = \frac{V^2}{P_{R_1}} = \frac{(120 \text{ V})^2}{40.0 \text{ W}} = 360 \Omega.$$

We are asked to find R_2 and \mathcal{E} . Use the expression derived in part (c).

$$I_0 = 0.600 \text{ A so } \mathcal{E}/R_2 = 0.600 \text{ A}$$

$$i = 0.150 \text{ A when } t = 0.080 \text{ s, so } i = \frac{\mathcal{E}}{R_2} e^{-(R_1+R_2)t/L} \text{ gives } 0.150 \text{ A} = (0.600 \text{ A}) e^{-(R_1+R_2)t/L}$$

$$\frac{1}{4} = e^{-(R_1+R_2)t/L} \text{ so } \ln 4 = (R_1 + R_2)t/L$$

$$R_2 = \frac{L \ln 4}{t} - R_1 = \frac{(22.0 \text{ H}) \ln 4}{0.080 \text{ s}} - 360 \Omega = 381.2 \Omega - 360 \Omega = 21.2 \Omega$$

$$\text{Then } \mathcal{E} = (0.600 \text{ A}) R_2 = (0.600 \text{ A})(21.2 \Omega) = 12.7 \text{ V.}$$

(e) IDENTIFY and SET UP: Use the expressions derived in part (a).

$$\text{EXECUTE: The current through the light bulb before the switch is opened is } i_1 = \frac{\mathcal{E}}{R_1} = \frac{12.7 \text{ V}}{360 \Omega} = 0.0353 \text{ A}$$

EVALUATE: When the switch is opened the current through the light bulb jumps from 0.0353 A to 0.600 A. Since the electrical power dissipated in the bulb (brightness) depend on i^2 , the bulb suddenly becomes much brighter.

30.76. IDENTIFY: Follow the steps specified in the problem.

SET UP: The current in an inductor does not change abruptly.

EXECUTE: (a) Using Kirchhoff's loop rule on the left and right branches:

$$\text{Left: } \mathcal{E} - (i_1 + i_2)R - L \frac{di_1}{dt} = 0 \Rightarrow R(i_1 + i_2) + L \frac{di_1}{dt} = \mathcal{E}.$$

$$\text{Right: } \mathcal{E} - (i_1 + i_2)R - \frac{q_2}{C} = 0 \Rightarrow R(i_1 + i_2) + \frac{q_2}{C} = \mathcal{E}.$$

(b) Initially, with the switch just closed, $i_1 = 0$, $i_2 = \frac{\mathcal{E}}{R}$ and $q_2 = 0$.

(c) The substitution of the solutions into the circuit equations to show that they satisfy the equations is a somewhat tedious exercise but straightforward exercise. We will show that the initial conditions are satisfied:

$$\text{At } t = 0, q_2 = \frac{\mathcal{E}}{\omega R} e^{-\beta t} \sin(\omega t) = \frac{\mathcal{E}}{\omega R} \sin(0) = 0.$$

$$i_1(t) = \frac{\mathcal{E}}{R} (1 - e^{-\beta t} [(2\omega RC)^{-1} \sin(\omega t) + \cos(\omega t)]) \Rightarrow i_1(0) = \frac{\mathcal{E}}{R} (1 - [\cos(0)]) = 0.$$

(d) When does i_2 first equal zero? $\omega = \sqrt{\frac{1}{LC} - \frac{1}{(2RC)^2}} = 625 \text{ rad/s.}$

$$i_2(t) = 0 = \frac{\mathcal{E}}{R} e^{-\beta t} [-(2\omega RC)^{-1} \sin(\omega t) + \cos(\omega t)] \Rightarrow -(2\omega RC)^{-1} \tan(\omega t) + 1 = 0 \text{ and}$$

$$\tan(\omega t) = +2\omega RC = +2(625 \text{ rad/s})(400 \Omega)(2.00 \times 10^{-6} \text{ F}) = +1.00.$$

$$\omega t = \arctan(+1.00) = +0.785 \Rightarrow t = \frac{0.785}{625 \text{ rad/s}} = 1.256 \times 10^{-3} \text{ s.}$$

EVALUATE: As $t \rightarrow \infty$, $i_1 \rightarrow \mathcal{E}/R$, $q_2 \rightarrow 0$ and $i_2 \rightarrow 0$.

30.77. IDENTIFY: Apply $L = \frac{N\Phi_B}{i}$ to calculate L .

SET UP: In the air the magnetic field is $B_{\text{Air}} = \frac{\mu_0 Ni}{W}$. In the liquid, $B_L = \frac{\mu Ni}{W}$

EXECUTE: (a) $\Phi_B = BA = B_L A_L + B_{\text{Air}} A_{\text{Air}} = \frac{\mu_0 Ni}{W} ((D-d)W) + \frac{K \mu_0 Ni}{W} (dW) = \mu_0 Ni [(D-d) + Kd]$.

$$L = \frac{N\Phi_B}{i} = \mu_0 N^2 [(D-d) + Kd] = L_0 - L_0 \frac{d}{D} + L_f \frac{d}{D} = L_0 + \left(\frac{L_f - L_0}{D} \right) d.$$

$$d = \left(\frac{L - L_0}{L_f - L_0} \right) D, \text{ where } L_0 = \mu_0 N^2 D, \text{ and } L_f = K \mu_0 N^2 D.$$

(b) Using $K = \chi_m + 1$ we can find the inductance for any height $L = L_0 \left(1 + \chi_m \frac{d}{D} \right)$.

Height of Fluid	Inductance of Liquid Oxygen	Inductance of Mercury
$d = D/4$	0.63024 H	0.63000 H
$d = D/2$	0.63048 H	0.62999 H
$d = 3D/4$	0.63072 H	0.62999 H
$d = D$	0.63096 H	0.62998 H

The values $\chi_m(\text{O}_2) = 1.52 \times 10^{-3}$ and $\chi_m(\text{Hg}) = -2.9 \times 10^{-5}$ have been used.

EVALUATE: (d) The volume gauge is much better for the liquid oxygen than the mercury because there is an easily detectable spread of values for the liquid oxygen, but not for the mercury.

30.78. IDENTIFY: The induced emf across the two coils is due to both the self-inductance of each and the mutual inductance of the pair of coils.

SET UP: The equivalent inductance is defined by $\mathcal{E} = L_{\text{eq}} \frac{di}{dt}$, where \mathcal{E} and i are the total emf and current across the combination.

EXECUTE: Series: $L_1 \frac{di_1}{dt} + L_2 \frac{di_2}{dt} + M_{21} \frac{di_1}{dt} + M_{12} \frac{di_2}{dt} \equiv L_{\text{eq}} \frac{di}{dt}$.

But $i = i_1 + i_2 \Rightarrow \frac{di}{dt} = \frac{di_1}{dt} + \frac{di_2}{dt}$ and $M_{12} = M_{21} \equiv M$, so $(L_1 + L_2 + 2M) \frac{di}{dt} = L_{\text{eq}} \frac{di}{dt}$ and $L_{\text{eq}} = L_1 + L_2 + 2M$.

Parallel: We have $L_1 \frac{di_1}{dt} + M_{12} \frac{di_2}{dt} = L_{\text{eq}} \frac{di}{dt}$ and $L_2 \frac{di_2}{dt} + M_{21} \frac{di_1}{dt} = L_{\text{eq}} \frac{di}{dt}$, with $\frac{di_1}{dt} + \frac{di_2}{dt} = \frac{di}{dt}$ and $M_{12} = M_{21} \equiv M$.

To simplify the algebra let $A = \frac{di_1}{dt}$, $B = \frac{di_2}{dt}$, and $C = \frac{di}{dt}$. So $L_1 A + MB = L_{\text{eq}} C$, $L_2 B + MA = L_{\text{eq}} C$, $A + B = C$. Now

solve for A and B in terms of C . $(L_1 - M)A + (M - L_2)B = 0$ using $A = C - B$. $(L_1 - M)(C - B) + (M - L_2)B = 0$.

$(L_1 - M)C - (L_1 - M)B + (M - L_2)B = 0$. $(2M - L_1 - L_2)B = (M - L_1)C$ and $B = \frac{(M - L_1)}{(2M - L_1 - L_2)} C$.

But $A = C - B = C - \frac{(M - L_1)C}{(2M - L_1 - L_2)} = \frac{(2M - L_1 - L_2) - M + L_1}{(2M - L_1 - L_2)} C$, or $A = \frac{M - L_2}{2M - L_1 - L_2} C$. Substitute A in B back

into original equation:

$\frac{L_1(M - L_2)C}{2M - L_1 - L_2} + \frac{M(M - L_1)}{(2M - L_1 - L_2)} C = L_{\text{eq}} C$ and $\frac{M^2 - L_1 L_2}{2M - L_1 - L_2} C = L_{\text{eq}} C$. Finally, $L_{\text{eq}} = \frac{L_1 L_2 - M^2}{L_1 + L_2 - 2M}$.

EVALUATE: If the flux of one coil doesn't pass through the other coil, so $M = 0$, then the results reduce to those of problem 30.47.

30.79. IDENTIFY: Apply Kirchoff's loop rule to the top and bottom branches of the circuit.

SET UP: Just after the switch is closed the current through the inductor is zero and the charge on the capacitor is zero.

EXECUTE: $\mathcal{E} - i_1 R_1 - L \frac{di_1}{dt} = 0 \Rightarrow i_1 = \frac{\mathcal{E}}{R_1} (1 - e^{-(R_1/L)t})$. $\mathcal{E} - i_2 R_2 - \frac{q_2}{C} = 0 \Rightarrow -\frac{di_2}{dt} R_2 - \frac{i_2}{C} = 0 \Rightarrow i_2 = \frac{\mathcal{E}}{R_2} e^{-(1/R_2 C)t}$.

$q_2 = \int_0^t i_2 dt' = -\frac{\mathcal{E}}{R_2} R_2 C e^{-(1/R_2 C)t'} \Big|_0^t = \mathcal{E} C (1 - e^{-(1/R_2 C)t})$.

(b) $i_1(0) = \frac{\mathcal{E}}{R_1} (1 - e^0) = 0$, $i_2 = \frac{\mathcal{E}}{R_2} e^0 = \frac{48.0 \text{ V}}{5000 \Omega} = 9.60 \times 10^{-3} \text{ A}$.

(c) As $t \rightarrow \infty$: $i_1(\infty) = \frac{\mathcal{E}}{R_1} (1 - e^{-\infty}) = \frac{\mathcal{E}}{R_1} = \frac{48.0 \text{ V}}{25.0 \Omega} = 1.92 \text{ A}$, $i_2 = \frac{\mathcal{E}}{R_2} e^{-\infty} = 0$. A good definition of a "long time" is

many time constants later.

(d) $i_1 = i_2 \Rightarrow \frac{\mathcal{E}}{R_1}(1 - e^{-(R_1/L)t}) = \frac{\mathcal{E}}{R_2}e^{-(1/R_2C)t} \Rightarrow (1 - e^{-(R_1/L)t}) = \frac{R_1}{R_2}e^{-(1/R_2C)t}$. Expanding the exponentials like

$e^x = 1 + x + \frac{x^2}{2} + \frac{x^3}{3!} + \dots$, we find: $\frac{R_1}{L}t - \frac{1}{2}\left(\frac{R_1}{L}\right)^2 t^2 + \dots = \frac{R_1}{R_2}\left(1 - \frac{t}{RC} + \frac{t^2}{2R^2C^2} - \dots\right)$ and

$t\left(\frac{R_1}{L} + \frac{R_1}{R_2^2C}\right) + O(t^2) + \dots = \frac{R_1}{R_2}$, if we have assumed that $t \ll 1$. Therefore:

$$t \approx \frac{1}{R_2}\left(\frac{1}{(1/L) + (1/R_2^2C)}\right) = \left(\frac{LR_2C}{L + R_2^2C}\right) = \left(\frac{(8.0 \text{ H})(5000 \Omega)(2.0 \times 10^{-5} \text{ F})}{8.0 \text{ H} + (5000 \Omega)^2(2.0 \times 10^{-5} \text{ F})}\right) = 1.6 \times 10^{-3} \text{ s}.$$

(e) At $t = 1.57 \times 10^{-3} \text{ s}$: $i_1 = \frac{\mathcal{E}}{R_1}(1 - e^{-(R_1/L)t}) = \frac{48 \text{ V}}{25 \Omega}(1 - e^{-(25/8)t}) = 9.4 \times 10^{-3} \text{ A}$.

(f) We want to know when the current is half its final value. We note that the current i_2 is very small to begin with, and just gets smaller, so we ignore it and find:

$$i_{1/2} = 0.960 \text{ A} = i_1 = \frac{\mathcal{E}}{R_1}(1 - e^{-(R_1/L)t}) = (1.92 \text{ A})(1 - e^{-(R_1/L)t}). \quad e^{-(R_1/L)t} = 0.500 \Rightarrow t = \frac{L}{R_1} \ln(0.5) = \frac{8.0 \text{ H}}{25 \Omega} \ln(0.5) = 0.22 \text{ s}.$$

EVALUATE: i_1 is initially zero and rises to a final value of 1.92 A. i_2 is initially 9.60 mA and falls to zero, q_2 is initially zero and rises to $q_2 = \mathcal{E}C = 960 \mu\text{C}$.